

A TEST AND ANALYSIS OF THE

SYNCHRONOUS MOTOR GENERATOR SET NO. M 4

IN THE ALTERNATING CURRENT LABORATORY

OF THE MICHIGAN STATE COLLEGE

THESIS FOR DEGREE OF B. S

W. D. HOWELL H. C. ROBERTS

1926

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A THESIS

SUBMITTED TO THE FACULTY

OF THE

MICHIGAN STATE COLLEGE

BY

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CANDIDATES FOR DEGREE OF

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### LIST OF SYMBOLS

- E = Impressed voltage.
- E' = Counter-voltage generated by motor.
- E\* = Resultant voltage.
- e = Power component of the impressed voltage.
- e' = Wattless component of the impressed voltage.
- e\* = Power component of the counter-voltage.
- e" = Wattless component of the counter-voltage.
- I = Armature current.
- i = Power component of the armature current.
- i' = Wattless component of the armature current.
- P = Mechanical power output of motor.
- r = Effective resistance of armature.
- x = Synchronous reactance of motor.
- z = Synchronous impedance of the motor.
- # = Angle between armature current and impressed
  voltage.
- e = Angle between armature current and resultant voltage.
- B = Angle between impressed voltage and countervoltage (reversed).

There has been a desire to have a separate source of Direct Current Power Supply in the Electrical Laboratories at Michigan State College which would be independent of the College Power Plant D.C. Supply.

This was accomplished recently by the installation of the Westinghouse Synchronous Motor-Generator Set.

This test was the first to be made on the M-G Set after installation in the Alternating Current Laberatory, and was performed with the view of obtaining its operating characteristics. The test as performed included not only the operating characteristics of the set, but also data to determine theoretical characteristics of the Synchronous Motor.

Throughout this test we have received excellent co-operation from the members of the Electrical Engineering Faculty and from Mr. F. Mitchell and Mr. Thompson of the Michigan State College Power Plant; for this we wish to express our sincere thanks.

### INTRODUCTION

The purpose of this test has been to obtain data for both practical and theoretical performance, and to supply experimental material which would be valuable to the department and to classes in Electrical Engineering.

Exercised to obtain accurate results as far as practicable. The measurements, in all cases, were taken by the use of suitable meters and transformers. The meters were calibrated after each test. Transformers were introduced in the high tension line in order to measure the power input by the instruments available and also to offer greater protection to those working around the instruments and machines. In this test the transformers were not calibrated because the calibration curves of these transformers showed but a slight variation in phase angle — less than five seconds.

ies, motor frames, control switches, and water barrel rheostats were grounded. This was done to give protection to those reading instruments or working with the apparatus. The water barrel rheostats used were made of metal and after experiencing several "pokes" it was decided to ground the barrels to the water main.

Advantage of the control of

When the meters were installed on the panels the field discharge resistance was lacking. An improvised resistance, consisting of a 220 wolt 60 watt bulb, was used but which exploded when the motor was started and an arc-over occurred in the compensator. After this experience it was decided to insert a resistance box with a rating of 1000 watts at 220 volts. This was found to work satisfactorily.

The power factor meter furnished would not operate to measure the power factor in the three phase system because it was a single phase meter and thus it does not measure the phase current and voltage. It was necessary to reverse the leads on the ammeters of the direct current generators, because, although the voltmeters would read in the same direction, the ammeters would read correctly when power was supplied to the machines but read back—wards when power was supplied by the machines. It was also necessary to take apart and readjust the circuit breaker on the 125 volt machine before it would function properly.

The synchronous motor bearing on the end toward the 250 volt generator was found to run hot throughout the test without apparent cause. The synchronous
motor was started by means of a compensator. This was
used in all cases when starting the motor instead of
trying to synchronize the machine with the supply line.

When starting, the motor drew a very heavy current in comparison with its name plate rating which amounted to about five and one-half times rated current at full load.

When running some of the tests certain peculiar phenomena occurred. This was expecially noticeable during the retardation runs and at certain times in the night which occurred generally in the form of static discharges or apparent grounds.

# CHAPTER I

THEORY OF THE SYNCHRONOUS MOTOR

## CHAPTER 1.

## THEORY OF THE SYNCHRONOUS MOTOR

## PRINCIPLE

The term "synchronous" means in unison, that is, in step. A synchronous motor then, may be defined as one which rotates in unison or in step with the phase of the alternating current which operates it.

A coil of wire if carrying current and placed in a magnetic field will tend to turn about an axis perpendicular to the field. This principle may be applied in the form of a motor to produce useful work. If such a coil is placed in a magnetic field as shown in Fig. 1

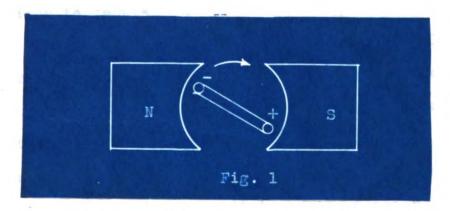


Diagram of an elementary motor.

and current allowed to flow through it, it can be seen by the left hand motor rule that the coil tends to turn in a clockwise direction. When the coil has reached the position as shown in Fig. 2, the coil would tend to turn counter-clockwise by use of the same motor rule.

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If the current in the coil was reversed just

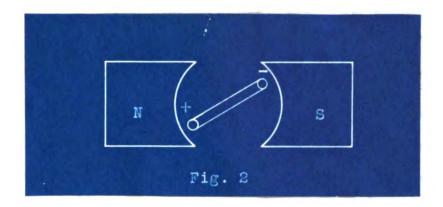


Diagram of an elementary motor
as the coil passes through its vertical or neutral
position, the turning moment of the coil would always be in the same direction and thus the coil would
continue to rotate.

This may be done by supplying the coil with alternating current and if the rotation of the coil keeps in synchronism with the current, the motor thus constructed will be a synchronous motor. Since action and reaction are always equal and opposite, it makes no difference whether the field or the armature revolves. On large machines it is generally found that the alternating current circuit is the stator and the field the rotor of the motor. Direct current is used for the field excitation.

A synchronous motor is in principle simply a reversed alternator. Instead of receiving mechanical energy and changing it to electrical energy, it receives electrical energy at synchronous speed and delivers mechanical energy. Synchronous motors are constructed as single and polyphase machines. This discussion will

treat only the poly-phase motor.

The condition necessary for synchronous motor operation is that the motor be speeded up until it rotates in synchronism, that is, in step with the alternator that supplies the power to run the motor. This means that the motor must run at the same frequency as the alternator but not necessary at the same speed. The speed of the motor is governed by the number of poles of the motor and the frequency of the power supply.

The principles upon which such motors operate may be explained by considering the action of an alternator and a synchronous motor connected in a circuit. Suppose that the motor is at rest when it is connected to the circuit with the alternator. The alternating current will flow through the armature of the motor and produce a reaction upon the field which would tend to rotate the rotor first in one direction then in the other. Because of the rapid reversal in direction of the torque thus set up, the time is not sufficient to allow the torque produced to overcome the inertia of the rotor, before a torque is produced in the opposite direction due to the reversal of the current. Hence the rotor remains stationary or, strictly speaking, vibrates. Thus, if the motor is first brought up to synchronism with the supply power and then connected in the circuit, the motor will continue to operate because the field flux and the armature flux are

always in the same relative position which produces a torque pulling the rotor around in the same direction.

The motor under test was started as an induction motor by means of an auxillary squirrel cage winding on the rotor. Reduced voltage was applied to the stator through a compensator and when the machine had reached approximately synchronous speed, the field switch was closed on the direct current circuit which brought the machine into synchronism after whach full voltage was applied to the stator. When the field switch is closed, the magnetic flux already set up in the fields by the armature reaction may be in the wrong direction in which case the rotor must fall back the distance of the polar pitch before the regular field flux can produce forward operating torque. Synchronous machines as compared to asynchronous machines run at a constant speed regardless of the load until such a load has been reached that the over-load capacity of the machine has been exceeded at which time the motor falls out of step and stops.

A synchronous motor has the important advantage that its power factor is adjustable and it can be made to take a leading or lagging current by varying the field excitation; its leading current can be used to offset the lagging current taken by induction motors thus improving the power factor.

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Synchronous motors are sometimes installed solely for this purpose.

### SYNCHRONOUS IMPEDANCE

Synchronous impedance may be defined as the ratio of the voltage that would be developed at a certain excitation, if the machine were run as a generator at synchronous speed without current in the armature, to the current that would flow through the armature at the same excitation when it is short-circuited. The synchronous impedance of the armature is composed of two components, i.e. effective resistance and synchronous reactance. Synchronous reactance includes reactance dum to induction and the effect of armature reaction in strengthening or weakening the field. These two effects are so similar that they are generally considered as one.

In The vector diagrams that are to follow the motor generated voltages are those that would be generated if the armature reactions were absent. To compensate for this assumption the synchronous impedance of the armature, which includes the effect of reactions, has been used instead of the ordinary impedance which consists only of true ohmic resistance and inductive reactance. The synchronous impedance has been considered constant with the variation of excitation, but in reality the field reactions and

position of the armature coils in reference to the field magnets and time phase of the current as well as upon the reluctance of the leakage magnetic circuit, so there is some error in assuming the synchronous impedance to be constant. Thus it is evident that the synchronous impedance is less for very high values of excitation than for low values, since the magnetic circuits are more highly saturated and a given armature magneto-motive force has less effect upon the field flux than for low values of excitation.

### VECTOR DIAGRAMS

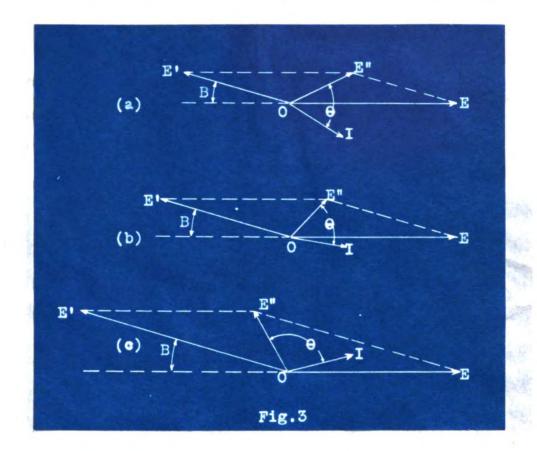
The complete action of a synchronous motor can be explained analytically by means of vector diagrams. Let E represent the impressed voltage, E' the counter voltage generated by the motor, and E' the resultant voltage that causes current to flow through the machine. Then E' equals E + E' and the armature current  $I = E^*/z$ , where z is the synchronous impedance. The current I lags behind E' by an angle  $\bullet$  whose tangent is x/r where x is the synchronous reactance and r is the effective resistance of the armature circuit.

In Fig. 3(a), suppose the motor is thrown on the line at a time when E' and E are exactly opposite in phase. E\* is then a minimum since,

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then it is the arithmetical difference between E and E'; the current I that flows is zero or so



(a) under-excited (b) under-excited (c) over-excited Fig. 3 Electromotive force diagram for a synchronous motor; three typical cases.

small and generally will not produce enough power to maintain rotation of the motor. The rotor tends to slow down, but as the rotor drops back in its running position E' drops back in phase, thus increasing E', I, and the power. A balance is reached at some point as d, when the power developed is just sufficient to meet the demand. The diagrams in Fig. 3 show three typical

cases where I lags behind, falls between, and preceeds E and E' prolonged.

The value of E' depends upon the excitation of the motor. If the excitation of the motor is
changed the resultant voltage and the synchronous
current, I, change in phase position and in magnitudes, but the motor still runs in synchronism.

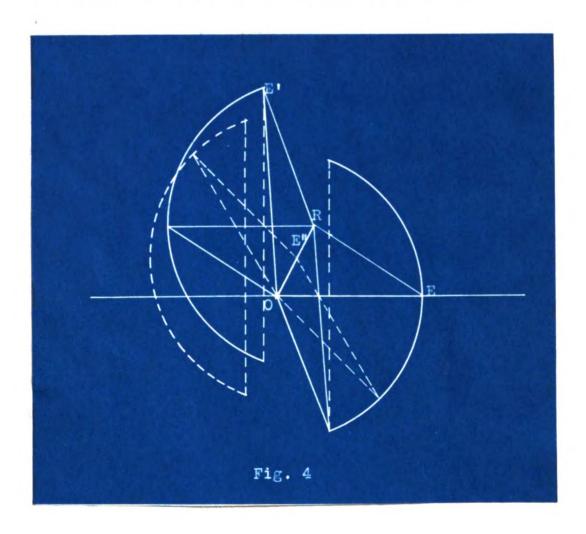


Fig. 4 The locus of the counter-voltage of a Synchronous motor.

The above Fig. 4 shows the locus of the

counter-voltage in the synchronous motor when the current and the impressed voltage and current remain constant. The current vector is made the reference vector. For changes in load the impressed voltage and counter-voltage vary in phase position and for any given excitation of the motor, the locus of the counter-voltage is a circle as shown above, with its center at R. The current I, is constant and also the resultant voltage E<sup>a</sup>, is constant if the synchronous impedance is assumed constant. The latter assumption is not exact because of the effect of saturation in the magnetic circuit with different excitations, as well as the effect of changes of power factor of the motor, but the assumption is sufficiently accurate for this explanation and test.

In order to have the resultant voltage E\* constant, the locus of the counter-voltage E' is a circle and must have its center at R and radius equal to E. This can be seen readily from the geometrical construction.

# CURRENT LOCI

Since the loci of the counter-voltage E' is a circle, the resultant voltage E', is also a circle because it is the vectorial sum of a circle and a constant; likewise the loci of the current will be a circle since it is equal to the resultant voltage, E',

Approximation of the control of the co

(whose locus is a circle) divided by the synchronous impedance. The center C is on the line OC making

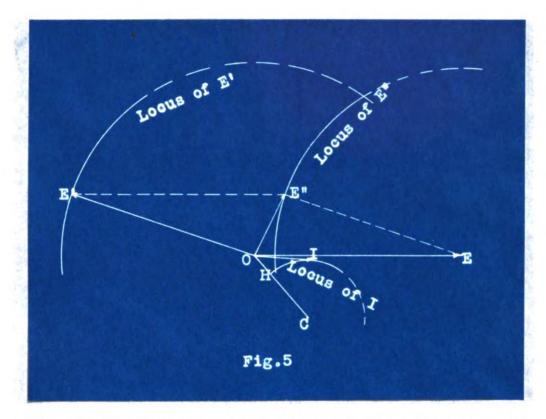


Fig. 5 Synchronous motor diagram; the solid part of each circle is the operating range.

an angle  $\theta$  with E (cosing  $\theta = r/z$ ); the length OC is equal to E/z, the radius CH is  $E^*/z$ . For different excitations the current loci consists of concentric circles with different radii determined by the relation CH;  $QC = E^* : E$ . The current locus passes through O when  $E^* = E$ . When the motor is underexecited the radius is less than OC and when overexecited the radius is greater than OC.

Fig. 6 is an expansion of Fig. 5 showing the current loci for a number of different excitations.

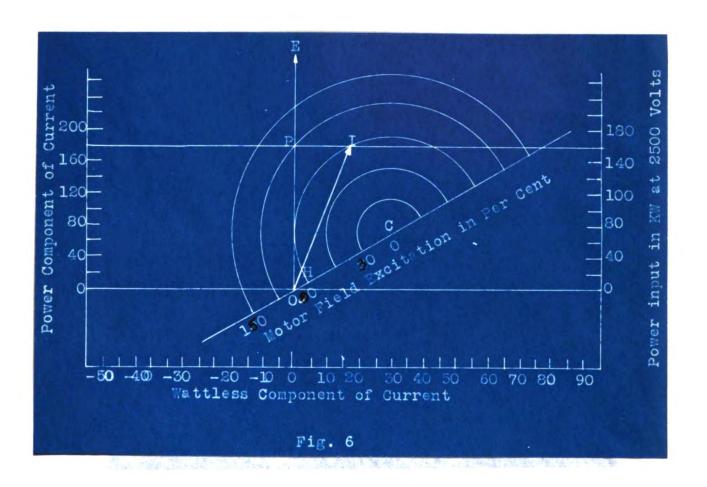


Fig. 6 Current loci for various motor excitations.

OI is the input current to the motor; OP is the power component of the current; OP/OI is the power factor = cosine Ø. OP x OE = I cos Ø E equals input watts; I E cos Ø - Losses = Output watts.

Efficiency = Output watts / Input watts. OC is the short-circuit current at full speed without

excitation; CH/OC = motor E.M.F./ supply E.M.F. = percent of excitation; OH is the minimum possible armature current.

## CIRCULAR CURRENT LOAD LOCI

at a given power factor to supply a certain load P, a locus of current for constant power may be determined from the above current loci.

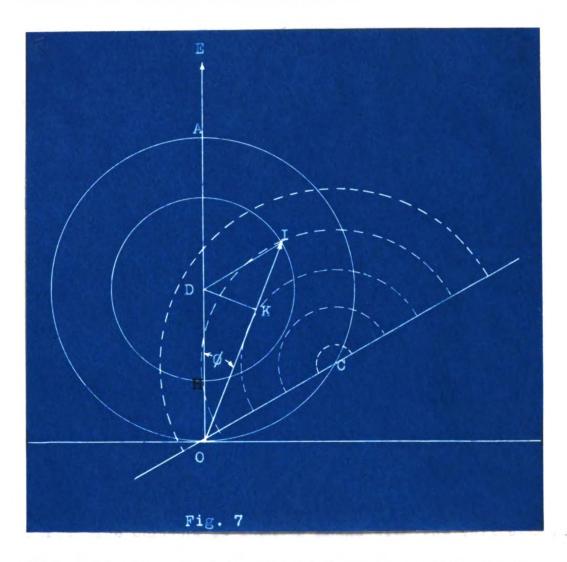


Fig. 7 Diagram showing current loci in a synchronous motor for constant power output when the impressed voltage is constant.

Thus in Fig. 7 let OE be the direction of the impressed voltage (which is made the reference vector). Assume that the rotor is delivering a certain output P including core and friction losses, then the input is

(1) E I cos  $\emptyset$  = P + I<sup>2</sup> r where E is the impressed voltage, I the current flowing, and  $\emptyset$  the angle of lag between the current and

impressed voltage. From this

(2)  $I^2 - \frac{E I \cos \emptyset}{r} = - P/r$  then completing the square

(3) 
$$I^2 - E \cos \phi I/r + \frac{1}{4} \frac{E^2 \cos^2 \phi}{r^2} = \frac{1}{4} \frac{E^2 \cos^3 \phi}{r^2} - \frac{P}{r}$$

therefore

(4) 
$$I = \frac{1}{2} \frac{E \cos \emptyset}{r} \pm \left(\frac{1}{4} \frac{E^2 \cos^2 \emptyset}{r^2} - \frac{P}{r}\right)^{\frac{1}{2}}$$

#### DETERMINATION OF THE RADIUS

From the above equation it is seen that there are two values of armature current for each value of cos Ø except one, and also that the currents are numerically equal but opposite in algebraic signs for equal positive and negative values of cos Ø.

From the equation for I it can be seen that the current I is equal to E/r when P and Ø are zero. The current is then in phase with the voltage, thus following Ohm's La w, making the loss entirely I<sup>2</sup>r.

This could only occur when the core losses and friction losses are supplied from external sources, and the rotor has taken such a position that the resultant voltage produces a current in phase with the inpressed voltage. This occurs when the counter-voltage is at right-angles to the current. This current, E/r, is laid off as OA in Fig. 7.

Take any value of current I for a given power P and for any angle  $\emptyset$  and assume I lags E by an angle  $\emptyset$  as drawn. Then draw the perpendicular DK from a point D midway between O and A and connect the point D and I. The length OD =  $\frac{1}{2}$ E/r equals  $\frac{1}{4}$  OA and by construction

(5) OK = 
$$\frac{1}{2}$$
 (E/r) cos  $\emptyset$ 

(6) 
$$DK = \frac{1}{2} (E/r) \sin \emptyset = \frac{1}{2} \frac{E}{r} (1 - \cos^2 \emptyset)^{\frac{1}{2}}$$
  
also (7)  $KI = OI - OK = I - \frac{1}{2} (E/r) \cos \emptyset$   
by construction

(a) 
$$(BI)^2 = (DK)^2 + (KI)^2$$

Substituting the values DK and KI and using the value for current from equation (4), then

(9) 
$$(DI)^2 = (\frac{1}{2} \frac{E}{r} - \sqrt{1 - \cos^2 \phi})^2 + \frac{1}{2} (E \cos \phi)/r$$
  
 $\pm (\frac{1}{2} \frac{E^2 \cos^2 \phi}{r^2} - P/r)^{\frac{1}{2}} - \frac{1}{2} (E \cos \phi)/r]^2$ 

which reduces to

(10) 
$$(\mathbf{D}I)^2 = \frac{1}{4} E^2/r^2 - P/r$$

or (11) DI = HD = 
$$\frac{+}{4} \left( \frac{1}{4} E^2/r^2 - P/r \right)^{\frac{1}{2}}$$

# 

Hence so long as the power P is constant, the locus of the current I will be a circle whose radius is equal to DI with its center D at a distance equal to  $\frac{1}{2}$  E/r from the origin O. For different ralues of power P the loci of the current are concentric circles with D as a center.

#### THE V AND O CURVES

The power factor of a synchronous motor is unity when the armsture current is a minimum for a constant mechanical output. By varying the field excitation of the motor the power factor may be varied with corresponding change in the armsture current.

The V curves show the relation between field excitation and the armsture current at a constant load and are usually plotted with field current as abscissa and armsture current as ordinates, or sometimes plotted with excitation an counter-volts as abscissa instead of field amperes. This constitutes the experimental method of determining the phase characteristics.

The V curves can be determined graphically as shown in Fig. 8. Find the intersections of the current locus, for a definite load, with the different field excitation loci. Plot on the graph the distances from the origin O to the intersections as ordinates against the corresponding field excitations as abscissa thus obtaining the V curves shown which are known as

the complete V curves or O curves. The heavy part of the curve is called the V curve and may be determined experimentally.

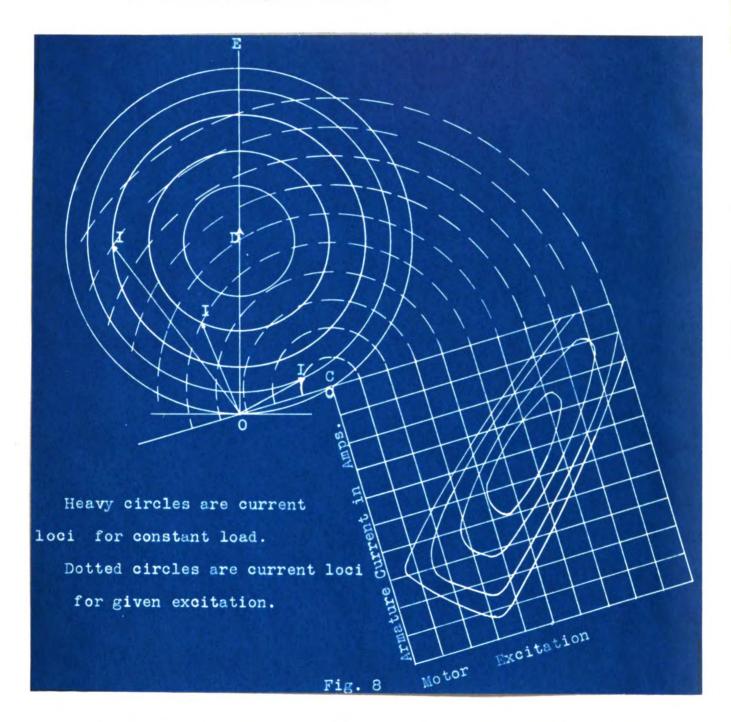


Fig. 8 Circular current loci; V curves and O curves.

# ANALYTICAL DETERMINATION OF THE PHASE CHARACTERISTICS

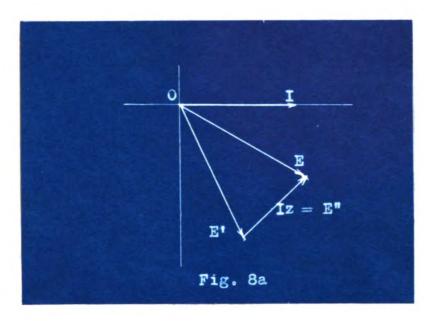


Fig. 8a Vector Diagram of voltages and current

Let E = impressed voltage on the motor.

E' = the voltage consumed by the counter-voltage of the motor.

Iz = drop or net loop voltage.

I = line current.

(12) z = r + j x =the complete impedance of a phase loop.

(13) 
$$I = i + j0$$

(14) 
$$E = e - j e^t$$

(15) 
$$E' = e'' - j e'''$$

Then 
$$(16)$$
 E = E' + Iz = e" + ir - j( e"' - ix)

• Therefore (17)  $e = e^{it} + ir$  and  $e^{it} = e^{it} - ix$ 

The motor power is  $P = e^{i}$  or  $e^{i} = P/i$ 

.

(18) 
$$E = (e^2 + e^{i^2})^{\frac{1}{2}}$$
 and  $E^i = (e^{i^2} + e^{i(i^2)})^{\frac{1}{2}}$ 

Eliminating e, et, em, emt, and solving for

i in the power equation

After squaring twice and transposing and collecting terms

(20) 
$$\mathbf{i}^{4} - \mathbf{i}^{2} \left[ \frac{2\mathbf{E}^{2}(\mathbf{x}^{2} - \mathbf{r}^{2}) + (\mathbf{r}^{2} + \mathbf{x}^{2})(2\mathbf{E}^{2} - 4\mathbf{Pr})}{(\mathbf{x}^{2} + \mathbf{r}^{2})^{2}} \right] + \left[ (\mathbf{E}^{2} + 2\mathbf{Pr} - \mathbf{E}^{2})^{2} + 4 \mathbf{P}^{2} \mathbf{x}^{2} \right] = 0$$

Which reduces to

(21) 
$$i^4 - B i^2 + C = 0$$

Hence by solving the quadratic equation of the second degree:

$$i = + \left[ \frac{B^{\pm} \sqrt{B^2 - 4C}}{2} \right]^{\frac{1}{2}}$$

# MAXIMUM OPERATING CONDITION

It can easily be seen from Fig. 9 that as the mechanical output increases the circle of the current locus decreases in size. Hence for maximum power output the circle of the current locus has contracted

The second of the second second of the secon

ented by OD. Under this condition one-half of the power input is used up on  $I^2r$  loss, and the other half in producing mechanical power, since  $P/r = \frac{1}{4}E^2/r^2$  and hence  $I = \frac{1}{2}$  E/r. The efficiency is thus fifty percent or less. If an arc is constructed with its origin at O and radius equal to OD, the intersection of this arc with the loci at various constant mechanical leads gives the maximum at which the motor will operate.

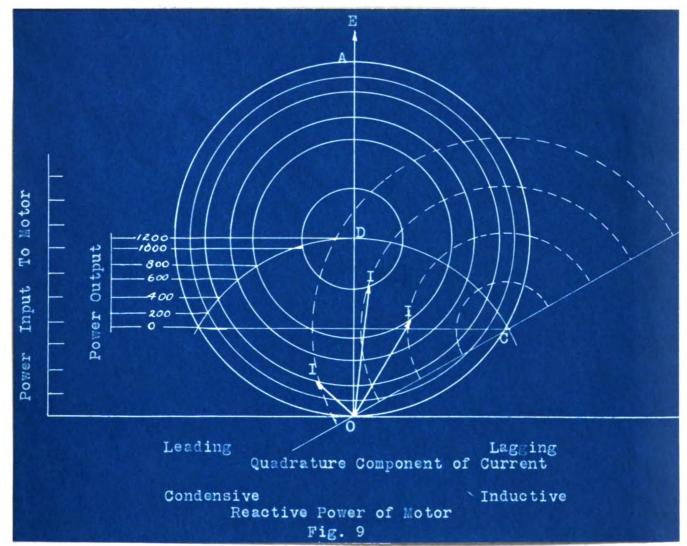


Fig. 9 Current for various mechanical power outputs.

MAXIMUM AND MINIMUM MOTOR EXCITATION FOR FIXED MOTOR
POWER AND FIXED IMPRESSED VOLTAGE

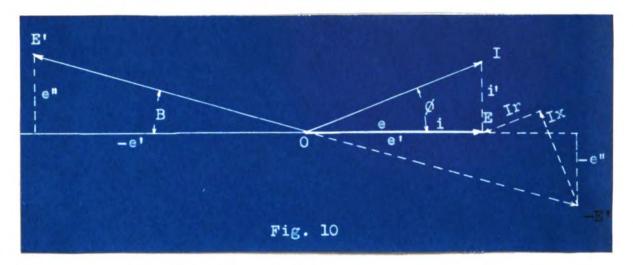


Fig. 10 Vector diagram.

(22) 
$$E = e + j0$$

(23) 
$$E^{\dagger} = e^{\dagger} + j e^{*}$$

(24) 
$$I = i + ji'$$

(25) 
$$E^* = E' - E = e' - je' - e$$

(26) I = E\*/z = (e'- je\* - e)/(r + jx)  
= 
$$\frac{(re' - re - xe*) + j(xe - xe' - re*)}{r^2 + x^2}$$

$$(27) = i + ji'$$

The power in any circuit is equal to the sum of the products of the real and imaginary parts of the voltage and current in respect to their order.

(28) 
$$P = e^i i + e^i i'$$

after clearing fractions:

(29) 
$$P = \frac{r E^{t^2} - e (r e^t + xe^t)}{r^2 + x^2}$$

The numerator of the expression for motor power of equation (29) for the condition of a fixed

 $\hat{\mathbf{y}} = \frac{1}{2} \left( \frac{1}{2} \left( \frac{1}{2} \right) \right) + \frac{1}{2} \left( \frac{1}{2} \left( \frac{1}{2} \right) \right)$ 

motor power, must be constant. From construction

(30) 
$$e^* = (E^2 - e^2)^{\frac{1}{2}}$$

Substituting in equation (29)

(31) 
$$P = \frac{rE^{2} - e(re^{1} + x - \sqrt{E^{2} - e^{2}})}{r^{2} + x^{2}}$$

The differential of P with respect to evwill be zero since the motor power is constant.

Differentiating:

(32) 
$$\frac{dP}{de!} = 2rE!\frac{dE!}{de!} - er - ex \left[ \frac{E!\frac{dE!}{de!} - e!}{-\sqrt{E!^2 - e!^2}} \right] = 0$$

Clearing fractions and solving for det

(33) 
$$\frac{dE^{\dagger}}{de^{\dagger}} = \frac{e\left[r - \frac{xe^{\dagger}}{(E^{\dagger}^{2} - e^{\dagger}^{2})^{\frac{1}{2}}}\right]}{2rE^{\dagger} - \frac{ex E^{\dagger}}{(E^{\dagger}^{2} - e^{\dagger}^{2})^{\frac{1}{2}}}} = 0$$
(34) Hence  $e\left[r - xe^{\dagger}/(E^{\dagger}^{2} - e^{\dagger}^{2})^{\frac{1}{2}}\right] = 0$ 

Transposing and squaring:

(35) 
$$r^2 E^{2} - r^2 e^{2} = x^2 e^2$$

Solving for e'

(36) 6 
$$e^{t} = (r E^{t}) / (r^{2} + x^{2}) = rE^{t} / z^{2}$$

Substituting the value of e and z in equation (31)

(37) 
$$Pz^2 = rE^{2} - er^2E^{2}/z - ex(E^{2} - r^2E^{2}/z^2)^{\frac{1}{2}}$$
  
=  $r E^{2} - ezE^{2}$ 

From which

(36) 
$$E^{*} = \frac{z}{2r}$$
 (  $e^{-\frac{1}{2}}\sqrt{4Pr + e^2}$  )

**:** 

But since E = e + j0
(39) E' = 
$$\frac{z}{2r}$$
 (E  $\frac{+}{2r}$   $\sqrt{4Pr + E^2}$ )

The maximum possible motor excitation comes when the motor power P equals zero, then the maximum excitation

$$(40) E^{\dagger} = Ez/r$$

The minimum excitation is zero.

MAXIMUM MOTOR POWER WITH FIXED EXCITATION, IMPRESSED VOLTAGE, EQUIVALENT RESISTANCE, AND SYNCHRONOUS IMPEDANCE

The motor power must be negative and therefore, since the first term of equation (29) is constant, the second must be negative in order to have maximum motor power. (re' + xe\*) must be a maximum to have maximum motor power.

Then:

(41) 
$$\frac{d}{de^{\dagger}}(re^{\dagger} + xe^{\bullet}) = r + x \frac{de^{\bullet}}{de^{\dagger}} = 0$$

(42) 
$$E^{2} = e^{2} + e^{2} = a constant.$$

$$(43) \quad \frac{dE^{\dagger}}{de^{\dagger}} = 2e^{\dagger} + 2e^{\bullet} \frac{de^{\bullet}}{de^{\dagger}} = 0$$

Combining equations (41) and (42) and (43)

$$\frac{d}{de!} (re! + xe!) = r - x \frac{e!}{e!} = 0$$

(45) Therefore 
$$r/x = e^{t}/e^{x}$$

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$$\frac{1}{2} \left( \frac{1}{2} + \frac{1}{2} \right) = \frac{1}{2} \left( \frac{1}{2} + \frac{1}{2} + \frac{1}{2} \right) = \frac{1}{2} \left( \frac{1}{2} + \frac{1}{$$

That is, the maximum power(motor) will occur when  $E^*$  lags behind E by an angle equal to  $\tan^{-1} x/r$ . Substituting the value of  $e^*$  from equation (45):

(46) Maximum 
$$P = \frac{rE^{2} - e(re^{2} + x^{2}e^{2}/r)}{z^{2}}$$

$$= rE^{2}/z^{2} - ee^{2}/r$$

But E = e + j0, therefore

(47) Maximum 
$$P = rE^{1/2}/z^2 - E e^{1/r}$$
  
And  $E^{1/2} = e^{1/2} + e^{1/2}$ 

And from equation (45)  $e^{x} = e^{x} x/r$ 

(48) Therefore 
$$e^{x^2} = E^{x^2} - e^{x^2}/r^2$$

Or: 
$$e^{z} = E^{z} r/z$$

Substituting this value of et in equation (47)

(49) Maximum power = 
$$E^{\dagger 2} r/z^2 - EE^{\dagger}/z$$

From equation (49) it follows that a synchronous motor with the excitation voltage greater than the impressed voltage cannot operate unless it has reactance. If the reactance is zero equation (49) becomes

(50) Maximum  $P = E^{1/2}/r - E^{1/r}$ For any value of E' greater than E, P will be positive and will represent generator action.

MAXIMUM POSSIBLE MOTOR EXCITATION WITH FIXED IMPRESSED VOLTAGE AND FIXED RESISTANCE AND RESCTANCE

P must be negative in order to have the

 $\frac{\partial \mathbf{g}}{\partial \mathbf{g}} = \mathbf{g} \cdot \mathbf{g} = \frac{\partial \mathbf{g}}{\partial \mathbf{g}} = \mathbf{g} \cdot \mathbf{g} \cdot \mathbf{g} = \mathbf{g} \cdot \mathbf{g} \cdot \mathbf{g} \cdot \mathbf{g} \cdot \mathbf{g} \cdot \mathbf{g} = \mathbf{g} \cdot \mathbf{$ 

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 $\{x_i\}_{i=1}^n$  ,  $x_i$  is the second of  $x_i$  and  $x_i$  is the second of  $x_i$  and  $x_i$ 

machine run as a motor. The limiting value of E' will be that value which makes P zero.

(51) Maximum  $P = E^{\dagger 2}r/z^2 - E E^{\dagger}/z = 0$ Therefore

(52) 
$$\frac{\mathbf{E}^{\dagger 2} \mathbf{r}}{\mathbf{z}^{2}} = \frac{\mathbf{E} \mathbf{E}^{\dagger}}{\mathbf{z}} \quad \text{or} \quad \mathbf{E}^{\dagger} = \mathbf{E} \mathbf{z}/\mathbf{r}$$

This equation shows that the maximum possible motor voltage is equal to the impressed voltage multiplied by the ratio of the synchronous impedance to the equivalent resistance.

MAXIMUM MOTOR ACTIVITY WITH FIXED IMPRESSED VOLTAGE
AND FIXED REACTANCE AND RESISTANCE

Substituting E = e + j0 in equation (51) and differentiating with respect to  $E^*$ :

(53) 
$$\frac{d\mathbf{P}}{d\mathbf{E}^{\dagger}} = \frac{\mathbf{d}}{d\mathbf{E}^{\dagger}} (\mathbf{E}^{\dagger}^{2}\mathbf{r}/\mathbf{z}^{2} - \mathbf{E}^{\dagger}\mathbf{e}/\mathbf{z}) = 0$$

Therefore

(54) 
$$2 E^{\dagger} r/z^2 = e/z$$
 or

(55) 
$$E^{\dagger} = e z/2r = E z/2r$$

The maximum motor power, therefore, occurrs when the motor voltage is one-half of its maximum possible value. Substituting the value of E' from equation (55) in equation (51), the maximum motor power then becomes:

(56) Maximum 
$$P = -E^2/4r$$

Equation (56) gives the maximum motor power possible.

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## OPERATING CONDITIONS OF A SYNCHRONOUS MOTOR

For the following discussion refer to Fig. 3. If E remains constant and E' is increased by increasing the field, the current I will move up to a leading position. If the excitation of the motor (E') is held constant but let E increase by putting a load on the motor, then as E increases, E' will also increase and follow a circle as a locus. The current I will also increase and depend upon E and shift to more lagging behind E. For a general case let E and E' be unequal; taking E as the reference vector let:

(57) 
$$E^* = E + E^{\dagger}$$
  

$$= E - E^{\dagger} (\cos B - j \sin B)$$

$$= E - E^{\dagger} \cos B + j E^{\dagger} \sin B$$

$$= e + j e^{\dagger}$$

Let the impedance of the loop circuit be z = r + jx which includes the resistance and reactance of the line and the machines.

(58) 
$$I = E^*/z = (E - E^* \cos B + j E^* \sin B)/(r + jx)$$

$$= \frac{1}{z^2} \left[ Er - E^* r \cos B + E^* \sin B + j (-E + E^* x \cos B + E^* r \sin B) \right]$$

$$= i + j i^*$$

Let P = motor power, then

(59)  $P = (E^* \sin B)i - (E^* \cos B) i^*$ 

From equation (57)

 $e = E - E^{\dagger} \cos B$  and  $e^{\dagger} = E^{\dagger} \sin B$ 

 $\mathbf{t}_{i} = \frac{\mathbf{t}_{i}}{2}$ 

 $||f_{i,j}||_{L^{\infty}} = ||f_{i,j}||_{L^{\infty}} = ||f_{i,j}||_{L^{\infty}} + ||f_{i,j}||_{L^{\infty}}$ 

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Then

(60) 
$$P = \frac{E^{\dagger}}{z^2} (e^{\dagger}r \sin B - ex \sin B - er \cos B - e^{\dagger}x \cos B)$$

Substituting the values for  $r = z \cos \theta$  and  $x = z \sin \theta$ :

(61) 
$$P = \frac{E'}{Z} (E' \cos \theta - E' \cos(\theta - B))$$

By differentiating P with respect to B, the maximum power is found for P when B =  $\Theta$ .

Substituting B =  $\Theta$  in the equation (61)

(62)  $P = \frac{E^{\dagger}}{2}$  (  $E^{\dagger}$  cos  $\Theta$  - E )

## HUNTING

Synchronous motors are subject to vibratory irregularities called hunting which are due to the fact that they depend upon a synchronizing current to hold them in step. The explanation of hunting can best be shown from a vector diagram as in Fig. 11.

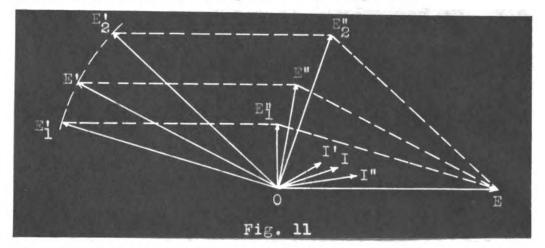


Fig. 11 Vector diagram showing the effects of hunting.

E is the impressed voltage and is considered

(0, 0, 0) = -100 + 100

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constant both in magnitude and phase position; E' is the counter-voltage and is constant in magnitude but not in phase position; E\* is the resultant voltage causing the current I in the armature and varies both in magnitude and phase position. I also varies in magnitude and phase position.

Now let the load on the motor be reduced to a lower value where the normal conditions would be represented by the voltage parallelogram and the current vector OI. However, these conditions would not be satisfied immediately because the speed of the rotor must increase slightly in order to reach this new position. Then, due to the inertia of the rotor, it would be carried on past the correct operating position so that the counter- voltage would take the position as shown by the vector  $OE^{\dagger}_{1}$ . This, of cenrse, will change the resultant voltage to CE\*, which has moved forward in phase position. The E\* will cause the current to take the new position as shown at OI'. Since the current has become less, the component of the current opposed to the counter-voltage is smaller and the resisting moment of the load being greater than the magnetic torque between the rotor and stator, the extra momentum is absorbed in the mechanical load and the

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moment of the load is still too great and the rotor must fall back in phase position. To accomplish this it must momentarily slow down in speed below synchronous. When the normal space—phase position has been reached the speed of the rotor is below normal and it drops back still farther, for instance, to the position OE\*2. The current also increases to OI\* as the rotor falls back and the magnetic torque is greater than the resisting moment of the load. The rotor again tends to speed up above synchronous speed. The vibrations described above may last for several cycles but finally die out due to the frictional, magnetic, and electrical losses which absorb some of the excess energy of the pendulum action of the rotor.

#### DAMPING

Hunting may be prevented to a certain extent by damping or by increasing the mass of the rotor with a fly-wheel. Damping may be accomplished by using a shortcircuited, low resistance winding on the field structure. These are sometimes called Amortisseur windings or dampers and are placed in the pole faces of the rotor. When hunting occurs, currents are induced in the Amortisseur winding which according to Lenz's Law oppose the flux producing them. Eddy currents and

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hysteresis losses are also produced in the pole faces which help slightly in the damping action. The induced currents in the damper oppose any change in the angular velocity of the rotor. For the most effective damping, the dampers should have as low resistance as possible but if this winding is to be used for starting the motor the resistance which gives the best damping may be too low to give the best starting torque.

#### STABILITY

Referring to equation (49) the

(63) Maximum  $P = (rE^2)/(r^2+x^2) - EE^2/(r^2+x^2)^{\frac{1}{2}}$ If we assume r, E,  $E^2$  fixed; then to obtain the value of x that will give the maximum power differentiate equation (63) with respect to x and equate the derivative to zero.

(64) 
$$\frac{dP}{dx} = \frac{-2xrE^{2}}{(r^{2} + x^{2})^{2}} + \frac{\frac{1}{2}(r^{2} + x^{2})^{\frac{1}{2}} 2 \times EE^{1}}{(r^{2} + x^{2})^{2}} = 0$$

Then

(65) 
$$-2 \text{ xrE'}^2 + (r^2 + x^2)^{\frac{1}{2}} \text{ xEE'} = 0$$

Solving for x2

(66) 
$$x^2 = r^2 \frac{4 E^{*2} - E^2}{E^2}$$

If E and E' are equal, then to give the maximum motor power

(67) 
$$x = r (3)^{\frac{1}{2}}$$
 or  $x/r = (3)^{\frac{1}{2}}$ 

Then from equation (67) the tan B =  $x/r = (3)^{\frac{1}{2}}$  and therefore this corresponds to a difference in phase between the impressed and excitation voltages of  $60^{\circ}$ . Breakdown will occur at this value of phase displaced ment. The counter-voltage will generally be equal to or greater than the impressed voltage as synchronous motors are used at unity or leading power factor. Therefore the ratio x/r should be equal to or greater than 1.732 to obtain maximum motor power. For maximum stability, the change in power developed should be a maximum for a given change in phase. In other words  $\frac{dP}{dB}$  should be a maximum. Referring to equation (29)

(68) 
$$P = \frac{rE^{2} - e(re^{1} + xe^{2})}{r^{2} + x^{2}}$$
$$= \frac{rE^{2} - EE^{1}(r \cos B + x \sin B)}{r^{2} + x^{2}}$$

Then differentiating, this expression becomes

(69) 
$$\frac{dP}{dB} = \frac{EE'(r \sin B - x \cos B)}{r^2 + x^2}$$

 $\frac{dP}{dB}$  is sometimes called the stability factor. It can be seen from equation (69) that the stability factor is directly proportional to E', the excitation voltage. A synchronous motor when over-excited is, therefore, more stable than one when operating under-excited. When tan B = x/r the motor developes maximum power. Then substituting the corresponding values



for sin B and cos B in equation (69),  $\frac{dP}{dB}$  becomes zero as it should. When tan B is greater than x/r,  $\frac{dP}{dB}$  is positive, and according to our notation (since motor power is negative) it represents a decrease in motor power. Therefore when tan B exceeds x/r the motor will fall out of step.

To make the stability factor a maximum, the relative magnitude of x and r can be found by equating the derivative of  $\frac{dP}{dB}$  with respect to x to zero.

(70) 
$$\frac{d}{dx} \left( \frac{dP}{dB} \right) = EE \cdot \frac{2x^2 \cos B - (r^2 + x^2) \cos B - 2xr \sin B}{(r^2 + x^2)^2}$$
  
= 0

Then simplifying

(71)  $x^2 \cos B - r^2 \cos B - 2xr \sin B = 0$ Hence

(72) 
$$x/r = \tan B + (1 + \tan^2 B)^{\frac{1}{2}}$$

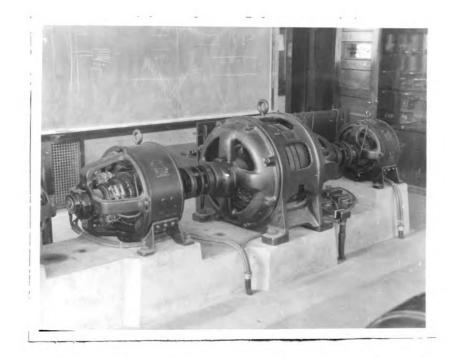
Since x can never be negative, the minus sign before the radical  $(1 + \tan^2 B)^{\frac{1}{2}}$  has no meaning. If the value of tan B in equation (72) is greater than or equal to x/r the condition of operation is unstable. For maximum stability the ratio of x/r is unity when B = 0, but this condition can never be obtained because at no load  $E^*$  must fall back enough to compensate for the losses in the motor. As B increases, the ratio x/r also increases for the condition of maximum stability.

The damper windings will affect these conditions slightly making the ratio x/r greater than that derived above.

# CHAPTER II

METHOD OF PROCEEDURE

# CHAPTER II NAME PLATE DATA



Synchronous Motor-Generator Set M - 4

# Westinghouse Synchronous Motor No. 4507244

50 Horse Power	1200 R.P.M.		
2200 Volts	Exciting Amperes = 17		
14 Amperes	Exciting Volts = 125		
80 % Power Factor	100 % Load		
3 Phase	24 Hours Continuous		
60 Cycle	400 C. Temperature Mise.		

# Westinghouse Direct Current Generator No. 4507242

Compound	Wound	1200	R.P.M.
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I5 Kilowatts 100 % Load

125 Volts 24 Hours Continuous

120 Amperes 40° C. Temperature Rise

# Westinghouse Direct Current Generator No. 4507246

Compound Wound 1200 R.P.M.

20 Kilowatts 100 % Load

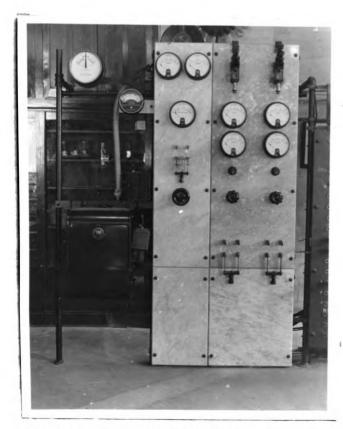
250 Volts 24 Hours Continuous

80 Amperes 40 C. Temperature Rise.

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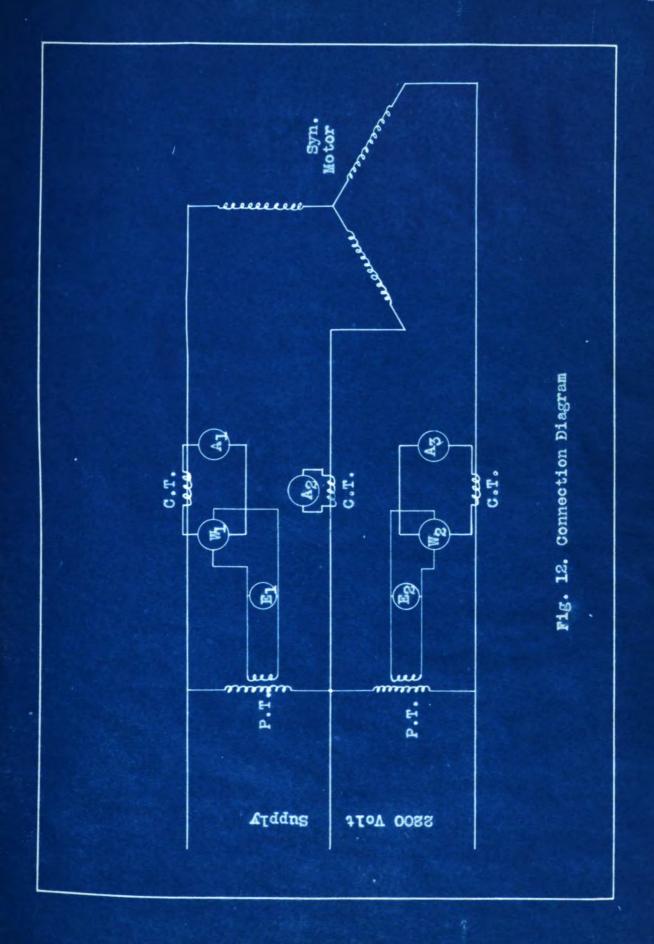
Instrument Transformer Connections



Panels of Motor-Generator Set M-4



Meter Connections



#### METHOD OF PROCEEDURE

#### GENERAL.

ed the operation of the machines, particular care was taken to see that the bearings were properly ciled and that the cil rings were functioning; also all meters were set at the zero reading and after the test was completed the meters were calibrated. A multiplier was used in conjunction with the potentiometer for calibrating the voltmeters and a standard shunt in conjunction with the potential shunt in

A speed counter was made from a small Direct Current fan motor driven as a generator with separate excitation. It was driven from the coupling on the shaft of the M-G set by a small belt. A voltmeter was connected across the armature terminals and calibrated by a precision tachometer to read in R.P.M.

#### STATIC TEST

The insulation of the three machines was first tested by means of a megger. The test consisted of measuring the approximate resistance between esperate parts of the machine and between different parts of the machine and ground. Measurements were

made from ground to slip rings, one end of field, one end of stator, other machines, conduit, one end of interpole winding, one end of shunt field winding, one end of series field winding, all segments on armature, and between commutator segments.

#### RESISTANCE MEASUREMENTS

The resistances of the different field windings and armature were found for the generators. Likewise measurements were made for the terminal resistance
of the stator windings and field winding of the synchronous motor. These measurements were determined by the
voltmeter-ammeter method in all cases, the voltmeter
measured the drop over the resistance only. The resistance of the voltmeter used was determined by the use
of the Wheatstone Bridge and standard resistances.
When the resistance measurements were made on the windings that had considerable inductance, the voltmeter
leads were disconnected before the current was interrupted.

#### DETERMINATION OF LOSSES

Losses were determined by the retardation method. The machine under test was run until the bearings were warmed sufficiently to make frictional losses constant. The generators were mechanically disconnected from the motor. Each generator was tested under the

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following conditions, i.e. the machine was brought up to rated speed as a motor and the power input recorded. Then it was speeded up above rating about one hundred R.P.M. at which time the power was shut off, the brushes raised immediately, and the armature allowed to come to rest. During this coasting period the instantaneous R.P.M. was recorded at equal short intervals of time. The above was repeated with brushes down and no field current; also with brushes down and with different values of field current.

chronous motor, it was driven by one of the generators operating as a motor. Tests were run on the synchron-ous motor for the same conditions as those used for the generators. The input to the driving motor at the rated R.P.M. was recorded in all cases, a constant value of field current was used on the driving motor. By knowing the losses and input of the driving motor, the losses in the synchronous motor could be determined.

For determining the efficiency of each machine the equation (Input - Losses) / (Input) was used. The losses were those determined as above in addition to the I<sup>2</sup>r losses of the machines.

#### SATURATION CURVES

The other generator was connected to the synchronous motor mechanically. The saturation curve of the synchronous motor was determined by driving it as

an alternator at the rated speed, by reading terminal volts at different values of field current and recording both. The field was demagnetized before running this test. Saturation curves were determined for the generators in a like manner.

#### TEST FOR SHAFT CURRENT

A high reading ammeter was connected with one end grounded to the frame and the other end bearing on the shaft through low resistance leads. If no current was noticed a smaller ammeter was used.

#### SYNCHRONOUS IMPEDANCE

The armature was shorted through suitable ammeters, first including the current transformers and second with the transformers cut out. The machine was driven at normal speed and sufficient field current applied to produce a good reading on the ammeters. Several readings were taken; for values of armature current from zero to 150 % of rating.

Attention was given to balancing of the phases at the same time. The synchronous impedance curve was plotted with field current as abscissa and armature current as ordinates.

### PHASE CHARACTERISTICS

Different mechanical loads were applied to the motor by loading the generators. V curves were taken for several different loads on the motor. The excitation was varied, for a definite load, from a very low value to approximately 160 % rated value. The armature current and field current were recorded for each variation. The wattmeter and voltage readings were also recorded. To determine the power factor, the ratio of the wattmeter readings was taken. V curves were plotted with field current as abscissa and armature current as ordinates.

#### LOAD VOLTAGE CURVES OF GENERATORS

Water barrel rheostats were used to load the generators. The generators were loaded by increments and readings of voltage and current taken at each increment. This was done at the same time that the V curves were obtained. Plotting load current at abscissa and terminal voltage as ordinates, the load voltage curves were determined.

#### STATIC TORQUE

An iron bar was fastened rigidly to the coupling on the shaft between the machines. This was used as a lever arm and a set of scales used to determine the pounds pressure exerted by the lever arm. The tare

was set permantly on the scales. For the various reduced voltages applied, readings of volts, line amperes, watts, and scale pounds were taken. By plotting the data obtained on logarithmic paper and the curve extrapolated for the starting voltage, the starting torque may be determined provided the function is a straight line. Another method is to construct the circle diagram corresponding to an induction motor from which approximate results may be obtained.

#### RUNNING TORQUE

The running torque was determined for the various loads by using equation

(73)  $T = \frac{7.04 \text{ Watts}}{\text{R.P.M.}}$  in pound-feet.

Output watts are used in the above equation.

#### PULL - OUT TORQUE

The pull out torque was determined by loading the generators to approximately three times
rated full load. At the instant of pull-out, readings
of watts, voltages, line amperes, and field amperes of
the motor were taken. The pull-out torque may be
determined from equation (73).

#### STARTING TESTS

The center of one pole on the roto r was aligned with the center of the frame. Then, two-thirds of the distance between pole centers was divided into five equal parts., On each one of these divisions the center-line of the marked pole was placed and the motor started under reduced voltage in each of the five positions. Readings were taken of volts, line amperes, and watts on the A.C. side.

<u>CAUTION:</u> Enough time was allowed between readings so as not to over-heat the machine and the current left on only long enough to obtain a reading.

#### WAVE FORM

The wave form of the voltage and current of the motor for all conditions from starting to running inclusive, were taken by means of the oscillograph.

### HUNTING AND DAMPING

For sudden changes of load within the rating of the motor no effects of hunting were noticed. It might be concluded from this that the damping was very effective.

CHAPTER III

DATA AND RESULTS

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### CHAPTER III

# DATA AND RESULTS

### Static Test

Evershed Megger Test Set No. 120661 500 volts, 1000 Megohms.

# Synchronous Motor No. 4507244

From	To	Megohms	
Ground on frame	Slip rings	1000 +	
	One end of field	Infinity -	
*	One end of stator	1000	
•	Other generators	O	
<b>u</b>	Conduit	0	

# D.C. Generator No. 4507242

From	To	Megohms	
Ground on frame	One end of Interpole field winding	1000	
10	One end of shunt field winding	Infinity-	
•	One end of seried field winding	Infinity-	
•	Armature	900	
	Commutator segments	900	

. .

D.C. Generator No. 4507246

From	To	Megohms
Ground on frame	One end of Inter- pole field	900
•	One end of shunt field	Infin <b>ity</b> -
•	One end of series field	Infinity-
•	Armature	900
•	Commutator segments	900

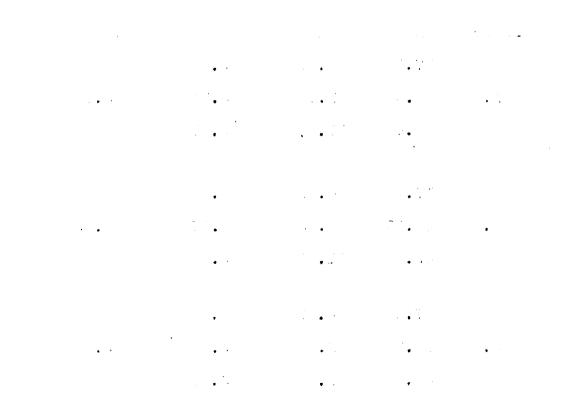
# RESISTANCE MEASUREMENTS

Synchronous Motor No. 4507244

Phase	E	I	R	Average R
	17.5	4.59	3.82	
ı.	30.8	8.03	3.74	3.79
	43.6	11.41	3.81	
	•			
	17.15	4.59	3.74	
2.	31.75	8.03	3.95	3.83
	43.6	11.41	3.81	
	17.6	4.59	3.82	
3.	30.7	8.03	3.74	3.81
	44.2	11.5	<b>3.</b> 88	

Average R of the three phases = 3.81 ohms





D.C. Generator No. 4507242

Part	E	I	R	Av. R
	132.9	2.84	43.3	
Snunt	65.2	1.40	45.7	
field	75.6	1.60	47.1	45 ohms
	64.2	1.40	43.3	
	43.3	•96	45.4	
(E in millivolts)	46.3	8,65	.00524	
Series	40.	7.58	•00528	
field in- cluding	33.9	6 <b>.3</b> 6	•0 <b>0533</b>	.00526 ohms
shunt.	26.9	5.07	.00530	
	19.1	3.70	.00515	
	•034	3.71	.00916	
Tu A our old o	.046	5.07	•00910	
Interpole field.	.057	6.39	•00898	.00908 ohms
	•068	7.59	•00909	
	1.30	2.66	<b>.4</b> 96	
Armature + Brushes.	1.64	4.04	.411	.415 ohms
	1.80	5.34	.339	
	.294	5.43	.0541	
Armature.	.417	7.73	•05 <b>4</b> 0	
	•220	4.04	•0545	.0541 ohms
	•154	2.66	.0579	
	•330	6.59	•0502	

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D	.C. Genera	tor No. 450	7246	
Part	E	I	R	Av. R
ah	114.2	.81	141.	
Shunt field.	116.8	.84	139.5	138.9 ohms
	130.1	•96	136.2	
	•035	3.64	.00967	
Series	.048	5.04	•00958	00056
field + shunt.	•060	6.35	•00952	•00956 ohms
	.07]	7.50	• <b>0094</b> 8	
	•022	1.37	.0157	
Interpole field.	.042	2.66	•0158	.0156 ohms
. I teru.	•056	3.60	•0154	
	1.80	<b>3.</b> 58	•5055	
Armature + Brushes	1.40	2.60	•5500	.5670 ohms
	0.80	1.36	•5960	
	•080	1.33	.0601	
Armature only	•180	2.60	.0694	.0648 ohms
	•230	<b>3.</b> 5 <b>5</b>	.0648	

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#### RETARDATION CURVES

Synchronous Motor No. 4507244

The following data contains the windage, bearing friction, brush friction, and eddy current losses of D.C. Generator No. 4507242 in addition to the heading losses of the test.

Windage of Syn.	notor	Windage + Brush F	riction'
Time in Min.	R.P.M.	Time in Mitt.	R.P.M.
0.0	1364	0.0 .05	1345 1281
<b>0</b> 5	1281	.10	1218
•1	1235		1153
.15	<b>11</b> 58	.15	
•2	1090	.20	1090
•25	1031	.25	1047
•3	983	• 30	990
	92 <b>7</b>	•35	951
<b>. 3</b> 5		•40	897
•4	85 <b>5</b>	•45	849
•45	820	• 50	803
<b>.</b> 5	770	•5 <b>5</b>	764
•55	727		719
•6	685	•60	
•65	642	<b>.</b> 65	685
•7	<b>\$</b> 10	•70	650
•75	578	•75	620
		•80	578
•8	5 <b>35</b>		

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The following data consists of the aforementioned losses of the D.C. Generator No. 4507242
and the windage, and brush friction of the synchronous motor, and the eddy current losses due to the
current rating flowing in the field winding as designated.

Time in Minutes

Field	Current = .5 amp.	Field Current = 1 amp	
Time	R.P.M.	Time R.P.M.	
0.0	1335	<b>0.</b> 0 <b>134</b> 5	
•05	1281	.05 1290	
.10	1230	.10 1240	
.15	1175	•15 1175	
.20	1111	.20 1111	
.25	1062	•25 <b>105</b> 8	
.30	1005	•30 1005	
<b>. 3</b> 5	962	•35 <b>9</b> 62	
•40	919	•40 913	
•45	862	• <b>4</b> 5 <b>855</b>	
•50	823	•50 813	
•5 <b>5</b>	787	•55 <b>77</b> 0	
•60	749	<b>.</b> 60 <b>73</b> 8	
•65	706	•65 706	
•70	663	•70 666	
Input	at 1200 R.P.M.	Input at 1200 R.P.	A .
#	2115 Watts	= 2270 Watts	

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Field	urrent = 3 amps	Field Cur	cent = 5 amps
Time	R.P.M.	Time	R.P.M.
0.0	1320	0.0	1326
•05	1260	•05	1260
.15	1183	•1	1175
.20	1111	•15	1090
•25	1047	•20	1025
• 30	98 <b>3</b>	•25	961
• 35	940	•30	898
•40	8 <b>76</b>	<b>⋄3</b> 5	8 <b>34</b>
•45	813	•40	7 <b>7</b> 0
•50	7 <b>7</b> 0	•45	706
•55	717	•50	663
•60	6 <b>74</b>	•55	620
•65	641	•60	57 <b>7</b>
•70	588	•65	535
•75	556	•70	481

Input at 1200 R.P.M = 2510 Watts

Input at 1200 R.P.M. = 2756 Watts

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10 (10 m) (10 m

Field	Current = 8 amps	Field Gurrent	= 10 amps
Time	R.P.M.	Time	R.P.M.
0.0	1370	0.0	1368
•05	1281	•O5	1281
.1	1174	•10	1163
.15	1068	.15	1048
.20	990	•20	958
.25	908	•25	845
• 30	834	• 30	749
• 35	748	• 35	663
•40	685	•40	598
•45	620	<b>.4</b> 5	513
•50	547	• 50	449
•55	492	•55	<b>3</b> 8 <b>5</b>
•60	<b>4</b> 28	•60	321
•65	374	•65	257
•70	<b>33</b> 2	•70	214
Inpu	t at 1200 R.P.M.	Input at	1200 R.P.M.
=	3380 Watts	= 3810	) Watts

Field Cu	rrent = 12 amps	jeld Curr	ent = 14 amps
Time	R.P.M.	Time	R.P.M.
0.0	1347	0.0	1341
•05	1240	•05	1240
.10	1132	•10	1132
.15	1025	.15	1025

Field Current	= 12 amps	Field Current	= 14 amps
Time	R.P.M.	Time	R.P.M.
•20	919	•20	919
.25	<b>8</b> 13	.25	812
• 30	727	• 30	727
•35	637	<b>.3</b> 5	642
• 40	545	•40	577
<b>.4</b> 5	492	<b>,</b> 45	492
• 50	406	•50	428
•55	342	•55	364
•60	278	•60	29 <b>9</b>
•65	214	•65	246
•70	171	•70	192
Input at 12 = 4020	00 R.P.N. Watts.	Input at 1200 = 4000 Wat	R.P.M.
= 4020	Wa <b>t</b> t <b>s.</b>	Input at 1200 = 4000 Wat Field Curren	ts
= 4020	Wa <b>t</b> t <b>s.</b>	= 4000 Wat	ts
= 4020 Field Current	Wa <b>tts.</b> = 16 amps	= 4000 Wat	t = 17 amps
= 4020 Time	Watts. = 16 amps R.P.M.	= 4000 Wat Field Curren Time	t = 17 amps R.P.M. 1335
= 4020 Time  6.0	Watts. = 16 amps R.P.M. 1331	= 4000 Wat Field Curren Time 0.0	t = 17 amps R.P.M. 1335
= 4020 Time Time 6.0	Watts. = 16 amps R.P.M. 1331 1220	= 4000 Wat Field Curren Time 0.0	t = 17 amps R.P.M. 1335 1219
= 4020 Trime Time 6.0 .05	Watts. = 16 amps R.P.M. 1331 1220 1111	= 4000 Wat Field Curren Time 0.0 .05	t = 17 amps  R.P.M.  1335  1219  1090
= 4020 Field Current Time 6.0 .05 .10 .15	Watts. = 16 amps R.P.M. 1331 1220 1111 1005	= 4000 Wate Field Current Time  0.0  .05  .10  .15	t = 17 amps  R.P.M.  1335  1219  1090  984
= 4020 field Current Time 6.0 .05 .10 .15 .20	Watts. = 16 amps R.P.M. 1331 1220 1111 1005	= 4000 Wat Field Curren Time 0.0 .05 .10 .15	t = 17 amps  R.P.M.  1335  1219  1090  984  877
= 4020 Trime Time 6.0 .05 .10 .15 .20 .25	Watts. = 16 amps R.P.M. 1331 1220 1111 1005 897 802	= 4000 Wate Field Current Time 0.0 .05 .10 .15 .20 .25	t = 17 amps R.P.M. 1335 1219 1090 984 877 701
= 4020 Trime Time 6.0 .05 .10 .15 .20 .25 .30	Watts. = 16 amps R.P.M. 1331 1220 1111 1005 897 802 717	= 4000 Wate Field Current Time 0.0 .05 .10 .15 .20 .25 .30	t = 17 amps  R.P.M.  1335  1219  1090  984  877  791

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Field C	surremt = 16 amps.	Field	Current = 17 amps.
Time	R.P.M.	Time	<b>R.P.</b> M.
0.50	406	•50	385
•5 <b>5</b>	<b>34</b> 2	•55	321
•60	278	•60	257
•65	214	•65	192
Input a	t 1200 R.P.M.	Input	at 1200 R.P.M.
= 41	10 Watts	=	4115 Watts

Field	Current = 18 amps	Field Curre	ent = 20 amps.
Time	R.P.M.	Time	R.P.M.
0.0	1350	0.0	1325
•05	1240	•05	1196
.10	1130	.10	1069
•15	1005	.15	965
.20	897	.20	855
.25	802	.25	749
• 30	717	• 30	652
•35	627	•35	566
<b>.4</b> 0	545	•40	492
•45	<b>4</b> 70	<b>.4</b> 5	406
• 50	406	•50	<b>3</b> 42
.55	342	•55	267
•60	267	•60	203
•65	214	•65	<b>1</b> 50
Inpu	t at 1200 R.P.M.	Input at	1200 R.P.M.
= 4	L76 Watts.	= 442	0 Watts

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Direct Current Generator No. 4507242

In each test the time interval is in minutes.

Windage +	Bearing Friction	Windage + Bea Frictio	
Time	R.P.M.	Time	R.P.M.
0.0	<b>13</b> 80	0.0	1355
•05	1317	•05	1189
.10	1250	.10	1061
.15	1168	.15	955
•20	1104	.20	867
•25	1047	<b>.2</b> 5	764
•30	976	. 30	680
•35	941	• 35	595
• 40	86 <b>6</b>	•40	516
•45	824	•45	465
• 50	764	• 50	<b>3</b> 82
•55	722	•55	310
.60	680	•60	234
.65	638	•65	197
•70	603	•70	106
Field Cu	rrent = 0 amps	Field Curren	nt = 0 amps.

of Windage, Bearing friction, and Brush friction, in addition to the loss incurred due to the field current flowing and thus causing an eddy current loss.

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Field	Current = 1.11 amps	Field Curre	ent = 1.22 amps.
Time	R.P.M.	Time	R.P.M.
0.0	1380	6.0	1355
•05	1253	•05	1189
110	1103	•10	1061
•15	986	•15	935
•20	850	·20	807
•25	<b>74</b> 0	•25	680
• 30	616	• 30	573
•35	499	•35	<b>4</b> 68
•40	403	•40	356
•45	255	•45	2 <b>4</b> 8
•50	191	• 50	166
•55	85	<b>. 5</b> 5	64
•60	0	•60	0
Inpu	t at 1200 R.P.M.	Input at	1200 R.P.M.
=	421 Watts	= 443	Watts.

Field čurrent =	1.42 amps.	Field Current :	= 1.63 amps.
Time	R.P.M.	Time	R.P.M.
0.0	1294	0.0	1294
•05	1146	•05	1125
.10	998	.10	934
•15	828	.15	754
•20	701	•20	595
.25	569	•25	467

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Field Current	= 1.42 amps	Field Current	t = 1.63 amps
Time	R.P.M.	Time	R.P.M.
•30	4 <b>4</b> 6	•30	<b>34</b> 6
<b>. 3</b> 5	319	<b>₊</b> 35	217
•40	212	•40	85
•45	106	<b>.</b> 45	0
Input at 1200	R.P.M.	Input at 120	00 R.P.M.
= 495 Watt	as .	≖ 564 Wat	tts

Field Curre	ent = 1.73 amps	Field Gurre	nt = 1.84 amps.
Time	R.P.M.	Time	R.P.M.
<b>6.</b> 0	1300	0.0	1380
•05	1130	•05	1168
•10	938	.10	977
•15	769	.15	786
•20	619	.20	6 <b>4</b> 8
.25	<b>4</b> 58	.25	488
•30	3 <b>37</b>	<b>.3</b> 0	361
• 35	192	•35	212
•40	77	•40	85
Input at 1	200 R.P.M.	Input at	1200 R.P.M.
= 620	Watts	= 630 W	la <b>t</b> t <b>s</b>

Field Curre	nt = 1.94  amps	Field Cur	rent = 2.05 emps
Time	R.P.M.	Time	R.P.M.
0.0	1295	0.0	1339
•0 <b>5</b>	1082	•05	1147
•10	871	•10	956、
•15	<b>7</b> 02	<b>,15</b>	744
•20	532	•20	574
•25	383	<u>.</u> 25	404
.30	2 <b>55</b>	•30	2 <b>55</b>
<b>.3</b> 5	106	.3	128
Input at	1200 R.P.M.	Input at	1200 R.P.M.
= 640 W	a t.t.a	<b>=</b> 690	Watts

Field	urrent =	2/161	amps
Time	•	R.P.M.	•
0.0		1320	
•05		1170	
•10	·	<b>956</b>	
•15		744	
.20		553	
.25		383	
<b>.3</b> 0		255	
• <b>3</b> 5		106	

= 700 Watts

Input at 1200 R.P.M.

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Direct Current Generator No. 4507246

Time interval in minutes.

Windage & Bearin	g Friction	Windage, Bear. friction and Brush friction	
Time	R.P.M.	Time	R.P.M.
0.0	1342	0.0	1300
•05	1283	•O5	1163
•10	1223	.10	1000
•15	1162	•15	885
•20	1098	.20	776
.25	1028	•25	69 <b>4</b>
• 30	959	•30	623
<b>. 3</b> 5	928	<b>. 3</b> 5	552
•40	863	<b>,</b> 40	474
•45	816	•45	<b>4</b> 08
• 50	776	• 59	347
Field Current =	O amps.	Field Current =	0 amps.

The following data will include the Windage, Bearing and Brush friction losses as well as that loss caused by eddy currents due to the field current as stated.

Field urrent =	•48 amps	Field Gurrent =	.61 amps.
Time	R.P.M.	Time	R.P.M.
0.0	1315	0.0	1305
•05	1203	•05	1204

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Field Gurrent =	.48 amps	Field current	= .61 amps
Time	R.P.M.	Time	R.P.M.
•10	1110	•10	1120
•15	959	•15	960
•20	867	•20	8 <b>53</b>
•25	775	•25	735
•30	675	• <b>3</b> 0	633
• <b>3</b> 5	583	<b>.</b> 35	551
<b>.4</b> 0	490	•40	449
•45	408	<b>.4</b> 5	368
•50	343	• 50	298
Input at 1200 R.	P.M.	Input at 120	00 R.P.M.
= 592 Watts		= 653.7 V	<b>latts</b>

Field Current =	.7 amps.	Field Current =	= .86 amps
Time	R.P.M.	Time	R.P.M.
0.0	1325	0.0	1300
•05	1192	<b>.0</b> 5	1160
.10	1058	.10	1020
.15	915	.15	878
.20	796	•20	756
.25	673	•25	634
• 30	572	• 30	531
•35	469	• <b>3</b> 5	429
•40	368	<b>.4</b> 0	337
•45	286	•45	225
•50	204	•50	153

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Input at 1200 R.P.M. Input at 1200 R.P.M.

= 714.4 Watts = 800 Watts

ield Current = .92 amps Field Current = 1.03 amps				
Time	R.P.M.	Time	R.P.M.	
0.0	1330	0.0	1310	
•O5 ,	1182	•05	1162	
•10	1020	.10	1000	
.15	87 <b>7</b>	•15	837	
•20	<b>7</b> 35	•20	694	
•25	613	•25	562	
•30	490	• 30	437	
<b>.</b> 35	388	•35	323	
• 40	286	•40	214	
•45	200	<b>.4</b> 5	122	
Input at 1	200 R.P:M.	Input at	1200 R.P.M.	
= 900 W	atts	= 948	Wa <b>t</b> ts	

Field Current =	1.29 amps	Field Curren	t = 1.54 amps
Time	R.P.M.	Time	R.P.M.
0.0	1308	0.0	1330
•05	1122	•05	1122
.10	943	.10	925
.15	776	.15	735
•20	633	.20	572
<b>.</b> 25	470	•25	429
•30`	<b>34</b> 8	• 30	286

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Field Curr	cent = 1.29 amps	Field Curre	nt = 1.54 amps
Time	R.P.M.	Time	R.P.M.
•35	245	•35	163
•40	133	•40	61.3
Input at	1200 R.P.M.	Input at	1200 R.P.M.
= 1024	Watts	= 1180	Watts

Field Current =	1.81 amps	Field Current =	2.06 amps
Time	R.P.M.	Time	R.P.M.
0.0	1350	0.0	1302
•05	897	•05	1080
•10	694	•10	837
.15	511	•15	613
•20	347	•20	428
<b>£25</b>	225	<b>.</b> 25	296
• 30	97	•30	143
<b>. 3</b> 5	0	•35	20.4
Input at 1200 1	R.P.M.	Input at 120	O R.P.M.
# 1270 Watt	e e	= 1360 Wat	ts

# SATURATION CURVES

# Synchronous Motor No. 4507244

Field	Terminal	Field	Terninal
Current	Volts O	Current	Volts
2.8	824.4	14.0	2436
4.2	1176	15.05	2486
5.05	1402	16.05	2492
6.25	1592	17	2588
7.0	1740	18.08	2626
8.0	1910	19.1	2672
9.0	2042	20.1	2720
10.1	2134	21.1	2750
11.0	2210	22	2066
12.0	2314	23	2786
13.0	2372	24	2794

# Direct Current Generator No. 4507242

Field Current	Terminal Volts	Field Current	Terminal Volts
<b>0</b> 0	00.0	1.10	88
.226	18.5	1.25	97
<b>. 4</b> 60	<b>3</b> 8	1.47	111
.552	46	1.595	120
.727	60	1.775	129
.829	67.5	2.00	139
•9 <b>4</b> 5	7 <b>7</b>	2.24	149

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Direct Current Generator No. 4507246

Field Current	Terminal Volts	Field Current	Terminal V <b>ėl</b> ts
0.00	000.0	•87	202
.45 9	120	•93	215
.52 <b>9</b>	133	•98	224
.5 <b>6</b> 2	142	1.04	235
.617	152	1.18	247.5
.671	166	1.313	265
•75	182	1.38	2 <b>7</b> 0
.823	192	1.53	280
		1.63	288

### TEST FOR SHAFT CURRENT

I = 0 For this test no current was found to flow between the shaft and the ground on the frame of the motor.

#### SYNCHRONOUS IMPEDANCE

# Synchronous Motor No. 4507244

Field Current	Phase 1 Current	Phase 2 Current	Phase 3 Current		
9.48	21	21.8	21.2		
8.00	17.9	18.8	18.		
7.1	<b>15.</b> 85	16.6	16.		

Field Current	Phase #1 Current	Phase #2 Current	Phase #3 Current		
6.2	13.9	14.55	14		
5.45	12.2	13.05	12.3		
4.45	9,95	11.15	10.1		
3.5	8,00	9.00	8.05		
2.75	6,35	7.25	6.3		
2.0	5.00	5.83	4.95		

### PHASE CHARACTERISTICS OF SYNCHRONOUS MOTOR

Synchronous Motor No. 4507244

The following data has been obtained by experiment, using the generators for a means of obtaining a steady load.

Watts 0 Mechanical Load =

s)Ag Current.	-17000 1.0	-15100 2.0	-12200 3.8	-8100 6.0	<b>-4800</b> 8.0	-1800 10.0	+1300 12.0	4200 14.0	7800 16.0	10200 17.0	12000 18.0	14500 20.0
W1 (watts) W2	24000	21200	17300	13000	0006	0009	2500	008 -	2000	0009 -	- 7000	- 10000
пз	17.8	15.8	12.8	<b>8</b>	5.8	<b>3.</b> 8	9•	2.1	4.6	7.3	8.6	10.3
H	17.8	15.7	12.7	8	ည	<b>13</b>	φ	2.1	4.6	7.3	8.6	10.3
냅	18.1	16.1	13.1	9.1	<b>9</b>	<b>3</b> •5	1.0	0.8	4.0	8.9	& &	8.6
(전 (전)	2260	2260	2260	2280	2300	2300	2320	2360	2360	2300	2300	2320
Ħ.	2260	2260	2260	2280	2300	2300	2320	2360	2360	2300	2300	2320

•	• 1 ×	• (	•	• 1 c	: • ·	. ·			•	•			
	\$	( ) ( )	• •	\(\frac{\cdots}{2}\)	,	i	1	1		( )  	1	:	:
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	, <b>k</b>	•	•	1 s	· .	•	•	•	la	5 10 10 10 10 10 10 10 10 10 10	•	1.1	
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\$ -	i.	v.			ξ.		\ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \	\$ · ·	•	a <sup>r</sup>	(	<del>⊢,</del>	

		Ä	Wethanical	Load =	Z Kw.		
Мľ	E CO	<b>н</b> т	H	r2	W <sub>1</sub> (K	(KW). Wg	Field Current
2280	2280	16.8	16.4	16.4	24	-14.	0.83
2300	2300	13.3	13.1	13.1	1912	-10.8	4.0
2320	2320	12.5	10.0	10.0	16	-7.3	0.9
2330	2330	7.2	6.9	6.9	18	0.4-0	8 •0
2360	2360	₽•8	<b>4</b> .3	<b>4</b>	O	8	10.0
2280	2280	8,0	α α	α α	8 5	+2.3	12.0
2290	2290	04 04	4.0	4.0	0.0	8.0	14.0
2300	2300	5.7	0.9	0.9	22	12.	16.0
2310	2310	7.8	80 10	ω •	-5.0	13.8	18.0
2300	2300	2.6	10.0	10.2	<b>0.7</b>	16.0	20.0

.

	Field Current	οı	4	9	ω	10	21	14	16	18	80
	<b>≥</b>	-11.8	7.4	2.00	00.00	+3.1	7.0	10	15	18.2	12
Kw.	W <sub>1</sub> (Kw)	27.8	23,2	17.8	15.	12.	8.0	5.0	ហ	ය ය	4.8
Load = 11.3	ı3	17.4	13.5	8.6	7.2	5.0	3. 8	4°5	7.4	10.2	11.8
Mechanical	H	17.4	13.5	8	7.8	<b>1</b> 0	κο Φ	4.3	7.4	10.2	11.8
Mech	다	17.5	13.8	9.65	7.6	ភ ស	ю 0	4.0	7.0	<b>9.</b> 4	11,1
	편 (%	0822	2290	2280	2300	2310	2330	2340	2300	2260	2280
	텀	2280	2290	2280	2300	2310	2330	2340	2300	2260	2280

Mechanical Load = 18.3 Kw.

Field Current	હ્ય	4	ဖ	ω	10	13	14	16	18	02
<b>*</b>	9	2°-	+1.4	ហ	8.1	11.6	14.8	18	20.7	23
W <sub>1</sub> (Kw)	62	22	20.5	16.8	13.5	10.2	7.2	4.5	• 01	0
L <sub>3</sub>	17.8	13,3	10.3	7.4	<b>9</b>	5.7	<b>9</b>	7.8	916	11.2
r S	17.2	13,3	10.3	7.6	1.9	ນ ໝໍ	∾ •	7.8	9.6	11,1
Ҥ	17.5	13.5	10.4	7.7	6.1	5.7	<b>८</b> 9	7.5	4.0	10.8
된 않	2170	2180	2200	2220	2240	2260	2280	2300	2310	2340
El.	2170	2160	2200	2220	2240	2260	2280	2300	2310	2340

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25.5 29.5 11 15 11 15 8 5 5 5 W<sub>1</sub> (Kw) +3.8 7.1 10.5 14.9 17.8 21.1 -10 27.1 KW. 15.2 1.2.1 1.3.1 7.8 7.4 7.9 10.2 Load = 15° 2 18° 1 9° 6 7° 8 7° 8 10° 2 12° 5 Mechanical 15.4 12.1 9.6 7.8 7.6 8.9 퇸

Mechanical Load = 29.6 KW.

된	南 以	ų	H	$\mathbf{I}_{3}$	W <sub>1</sub> (Kw)	v) W2	Field Current
2160	2160	16.2	16	16,2	32	Q	4
2180	2180	13.3	13	13,2	27.5	<b>9</b>	ဖ
2180	2180	11.1	10.9	11	24.6	11	ω
2200	2200	10.9	10.7	10.8	<b>2</b> 4	10	10
2220	2220	و د	1.6	9.1	80	13.5	12
2200	2200	<b>ω</b>	8.6	8 • 6	17	16.8	14
2890	2290	• 6	თ	6.8	13.5	80	16
2300	2300	10	11	6.6	11.5	23	18
2330	2330	12:2	12.5	12.5	9	27.8	003

<b>គ</b> ្	MECHANICAL L, I.	LOAD =	34.2 KW.		Field
Ë	ĸ	C	, R H R		Current
<b>1</b>		14.8		1	۴ <b>د</b>
1 33	12.6 12.2		28 28	1045	ο დ
_			24.5	14	10
	9•6 9•5	0 •	20.5	17.9	13
	9.6 7.6		17.2	12	14
	10.6 10.6	5 10.5	24.8	24.6	16
			11.5	27.5	18
	13 13	13	10	30.1	80

Mechanical Load = 45.5 KW.

F,
14.2
12.7
21
12,3
13,1
13.8
14.8

EFFICIENCY OF SYNCHRONOUS MOTOR

Synchronous Motor No. 4507244

Input in KW.	<pre>Input - Losses in KW</pre>	Efficiency in Percent
5.8	2	<b>34.</b> 5
15	11,3	75.5
21.8	18.3	84
29.9	27 <b>.</b> x	90.5
<b>33.</b> 8	29,6	87.6
<b>3</b> 8.2	<b>34.</b> 2	89.5
48.3	44.3	91.8

Efficiency = Input - Losses
Input

Values above taken for unity power factor.

### LOAD-VOLTAGE CURVES OF GENERATORS

Direct Current Generator No. 4507242

Field Current = 1.43 amps

E	I	Watts	E	I	Watts
119	0	0	125	76	9500
125.5	16	2010	125	86	10740
124	29	3600	125	99	12370
123	46	5660	124	112	13800
124	65,5	8125	120	124	14900
·			120	138	16580

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<b>、</b> • · ·	•	•

Direct Current Generator No. 4507246

Field Current = 1.2h

K	I	Watts	E	I	Watts
258	0	٥	255	62	15800
256	17.5	4480	252	72	18120
258	29.5	7620	251	84	21050
257	39	10020	248	98	24520
25 <b>5</b>	50	12750			

#### STATIC TORQUE

Torque on Synchronous Motor No. 4507244

E <sub>1</sub>	E2	1,	ı	13	W <sub>1</sub> (Mil	2 (wo	Torque 1bft.
492.6	492.6	21.5	22	23.3	21,2	1.	22,5
554	554	23.6	24.7	26.2	24.6	1.6	33
605.4	605.4	25.8	27.1	29.5	30.4	3	51

Static torque taken with blocked rotor and reduced voltage. Tare on scales = 45.5 lbs. Torque measured with synchronous measure operating as an induction motor with rotor blocked.

CIRCLE DIAGRAM OF SYNCHRONOUS MOTOR OPERATING AS AN INDUCTION MOTOR

Motor running idle on starting position of compensator.

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By waing the data from the Static Test in conjunction with this given above, the circle-diagram may be consturcted.

$$I_0 = 15.5$$
 amps

$$\Theta = 63^{\circ} 16^{\circ}$$
 or P.F. = .45

$$\theta = 55^{\circ} 57^{\circ}$$

Primary copper loss =  $(78)^2 \times 2.191 = 13,350$  watts

Exciting watts at no load = 8000

Total loss at blocked rotor = 84,000 watts

Secondary copper loss = 62,650 \*

Thus the starting torque =  $7.04 \times 62.650$ 1200

= 368 lb. ft.

#### RUNNING TORQUE

Synchronous Motor No. 4507244

Torque derived from the equation

$$T = \frac{7.04 \times (\text{watts})}{\text{Syn. R.P.M.}}$$

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· · · · · ·

Load in Watts	Torque in 1bft.
2000	11.7
11,300	66.4
18,300	107.2
27,100	159
29,600	173.5
34,200	200
44,300	260

#### PULL-OUT TORQUE

Synchronius Motor No. 4507244

ield Gurrent = 14.5 amps

E	E	I <sub>1</sub>	ıs	<b>1</b> 3	$W_1$ (Kw)	w <sub>2</sub>
2000	2000	64	6 <b>6</b>	<b>6</b> 0	98	57
2000	2000	64	66	<b>6</b> 0	98	<b>57</b>

Unity Power Factor

Pull-out torque = 908 lbs.-ft. (maximum possible withous consideration of losses).

# STARTING TESTS ON SYNCHRONOUS MOTOR Conditions at starting compensator

Position	E	Gondition					
Q	2340	Open line voltage.					
1 st	2250	Starter on starting position.					
2 nd.	2340	Field put on (14.5 amps) with					
		auto on starting position.					

Position	E	Condition
3 rd.	2360	Auto starter in running position, no load on motor.
4 th.	2050	Supply line voltage when motor pulled out of step due to loading.

conditions at motor with Auto on starting position: (motor running as an induction motor)

E <sub>1</sub>	E	I <sub>1</sub>	$\mathbf{I}_{2}$	13	T <sub>1</sub> (Kw)	WS
1780	1780	71	<b>7</b> 3	76	94	9.4

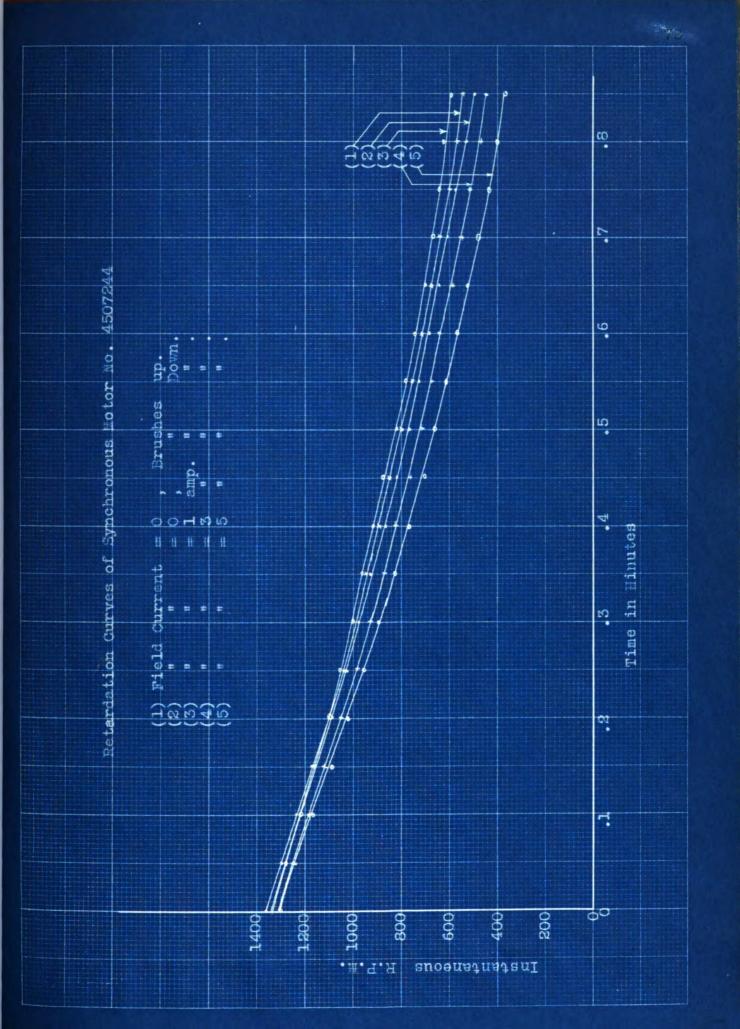
Starting tests with reduced voltage for the different positions of the field poles.

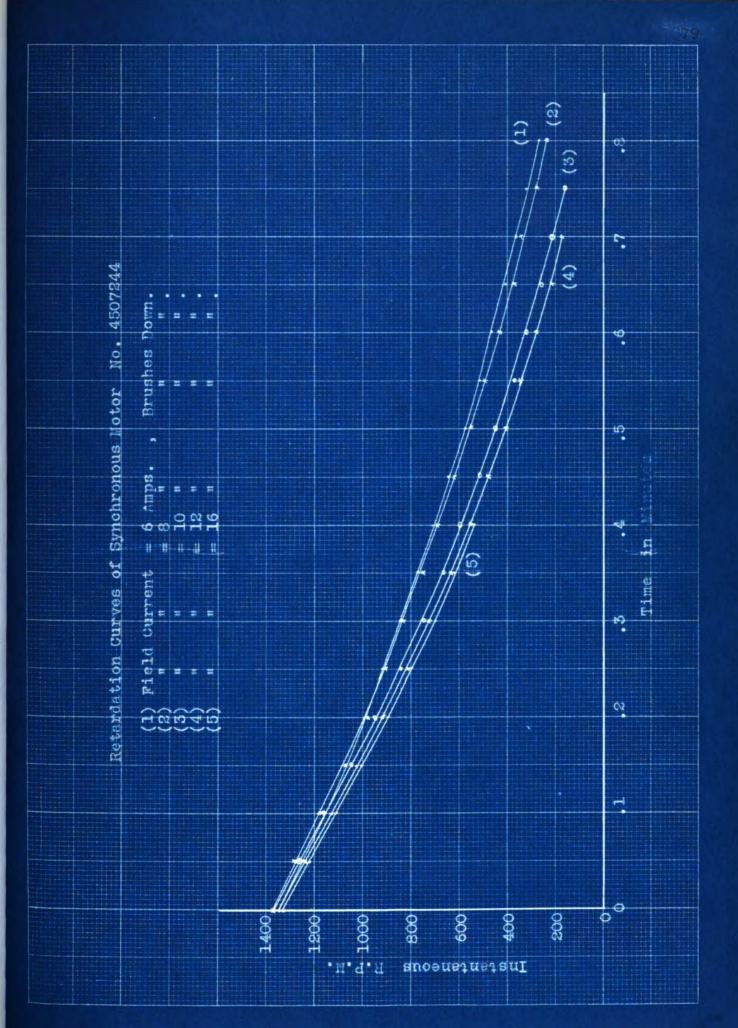
E	E2	11	ı	13	W <sub>1</sub> (Kw)	<b>W</b> 2	Field Current A.C.
500	<b>53</b> 0	24	23.5	21.3	10	1	3.4
500	520	23.4	24	21.5	10	2	3.4
450	550	20.4	21.2	20	7.4	1.6	3.06
<b>4</b> 80	440	20.1	21.6	20.5	8	1.2	3.1
500	440	20.5	21	22	8.1	1.2	3.12
520	460	21.5	20.5	22.5	9.8	1.4	3.22
4 <b>4</b> 0	480	21.5	20	19.5	9	1.8	3.05

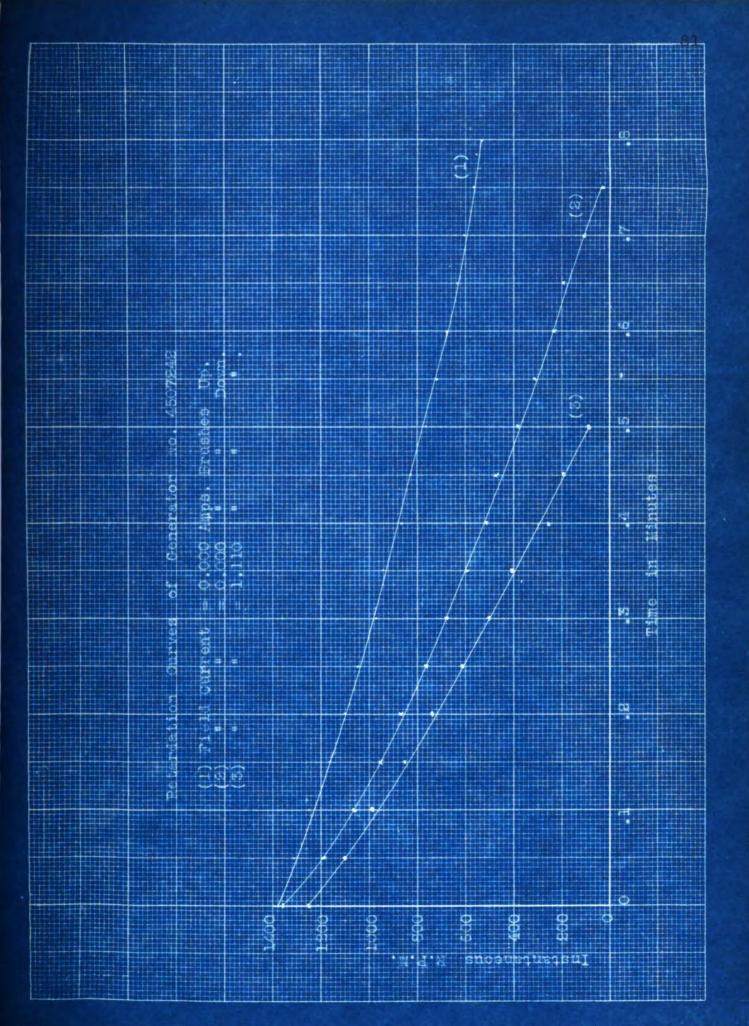
Field shorted through 48.4 ohms resistance.

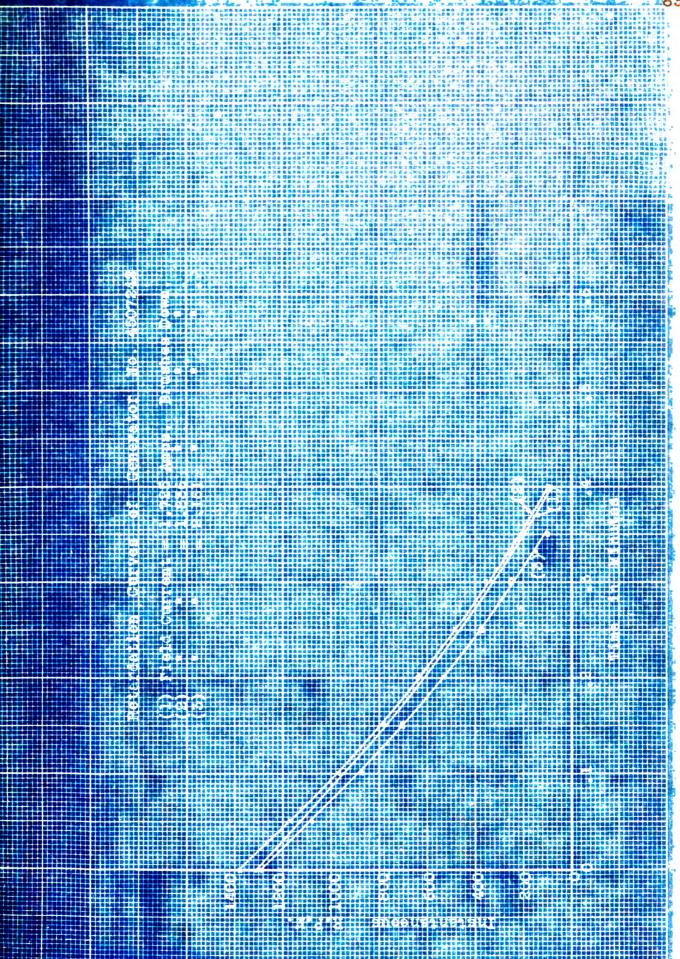
CHAPTER IV

GRAPHS



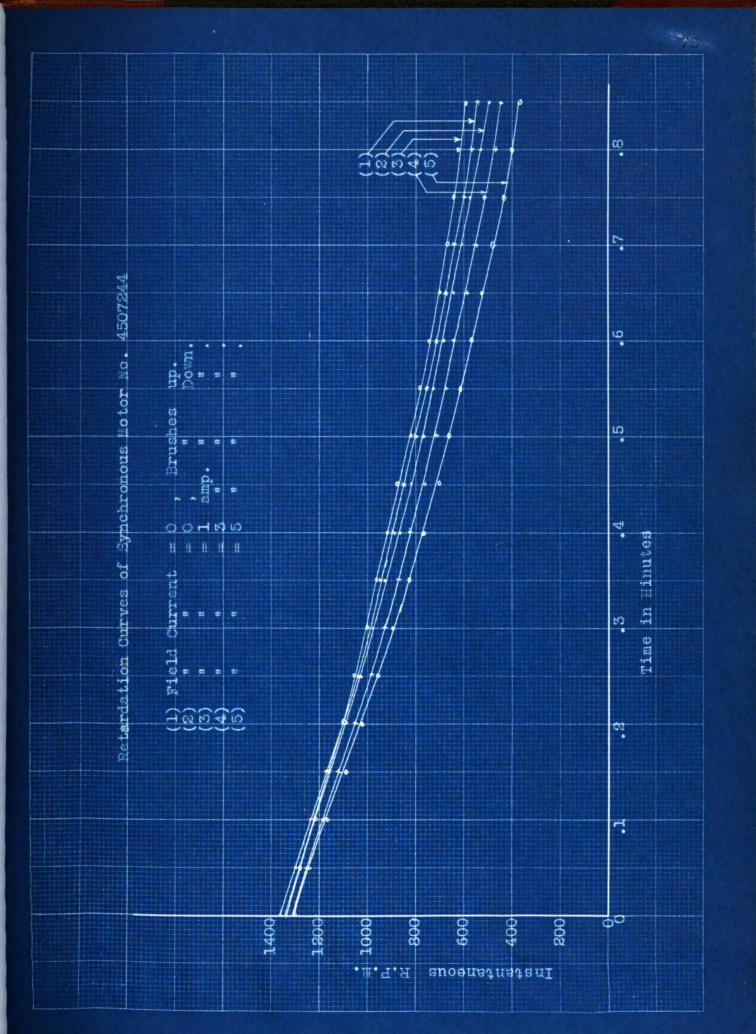


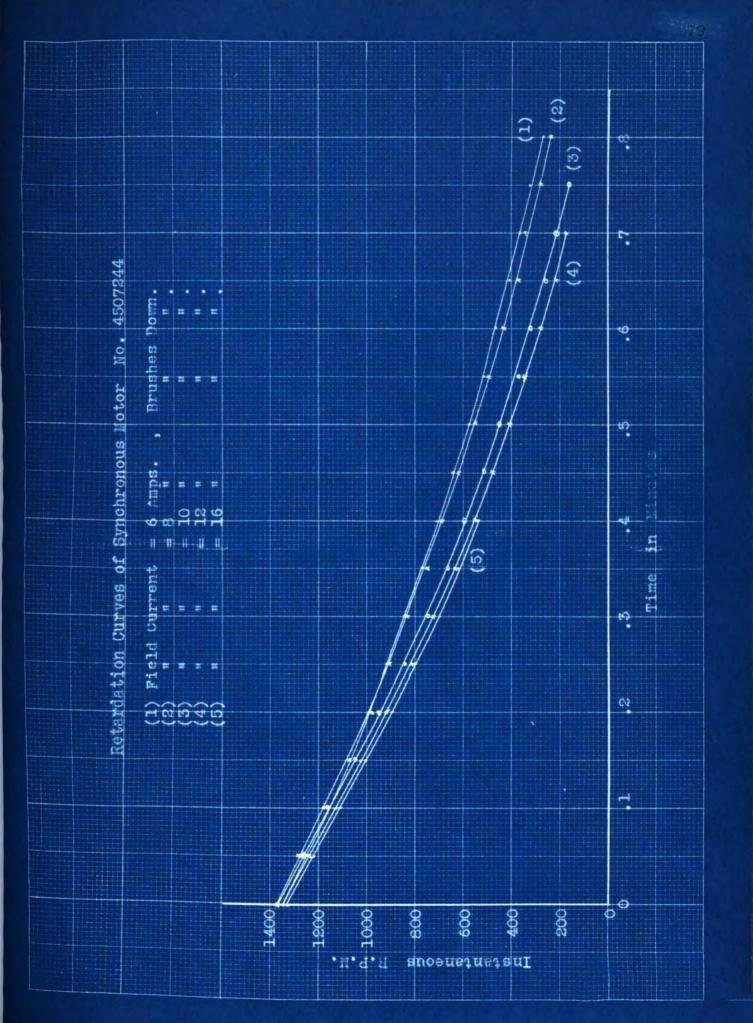


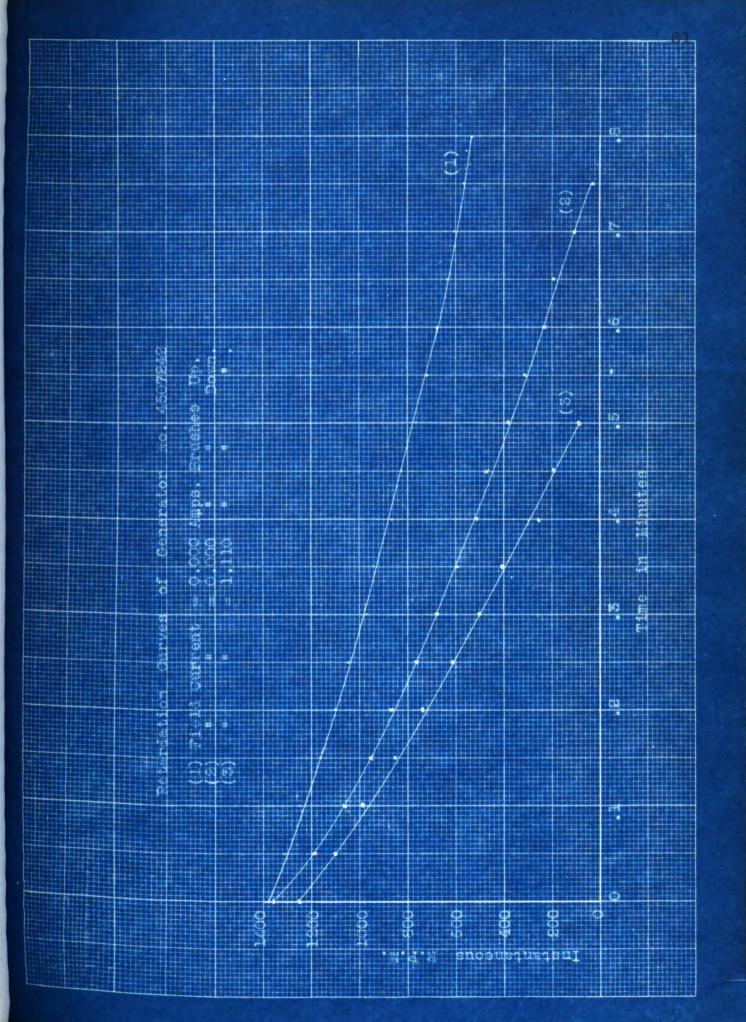


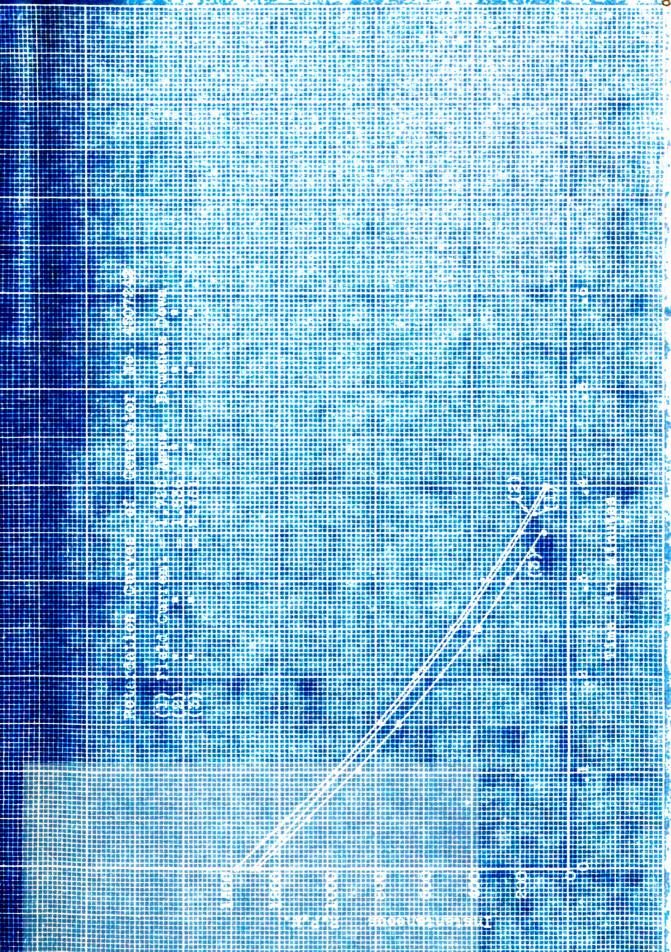
CHAPTER IV

GRAPHS



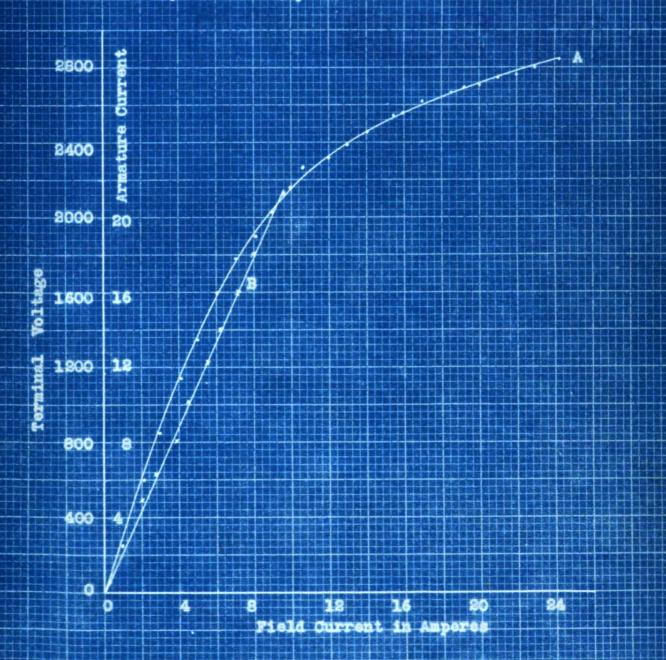




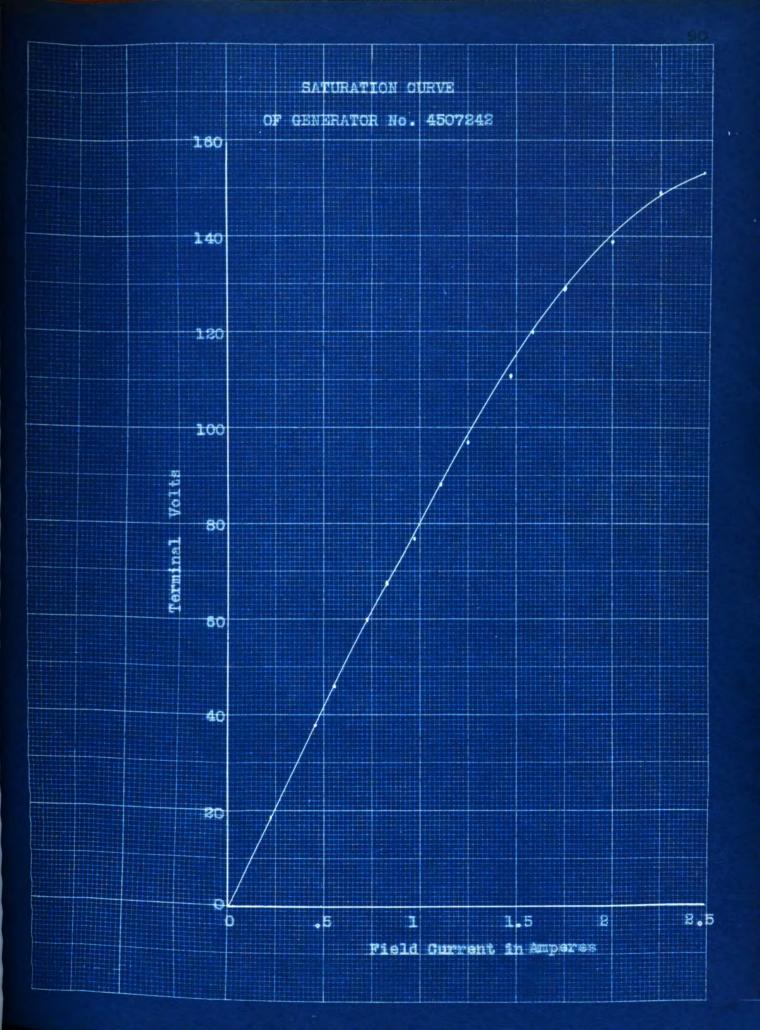


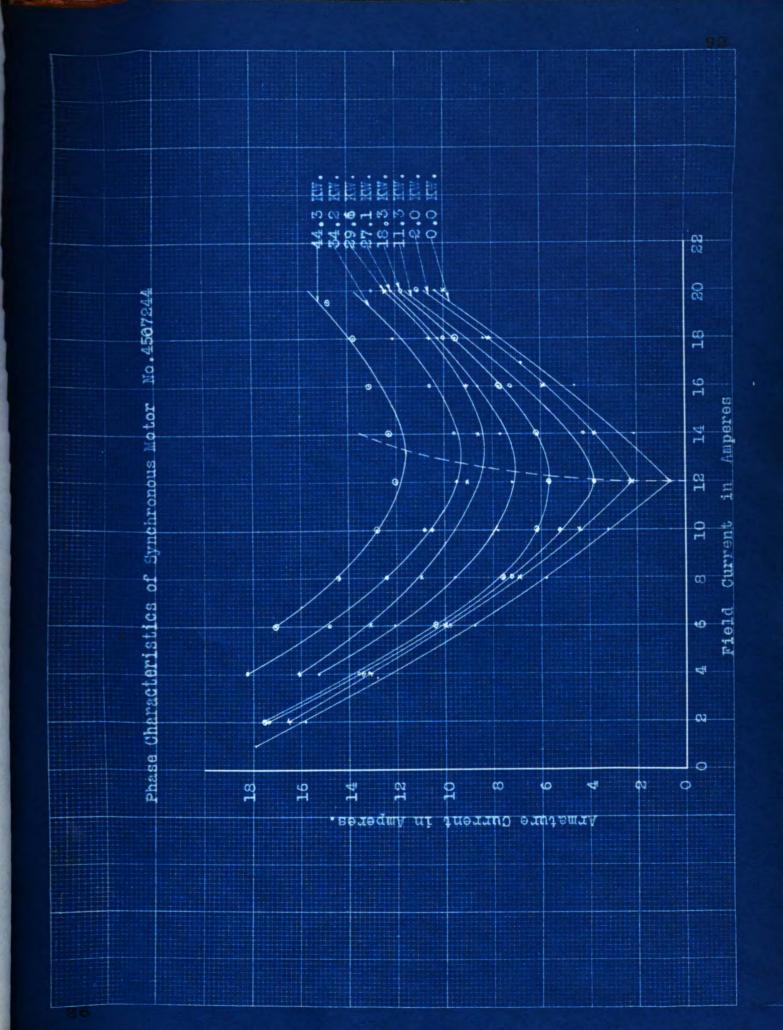
## SATURATION CURVE AND SYNCHRONOUS IMPEDANCE CURVE OF SYNCHRONOUS MOTOR No. 4507244

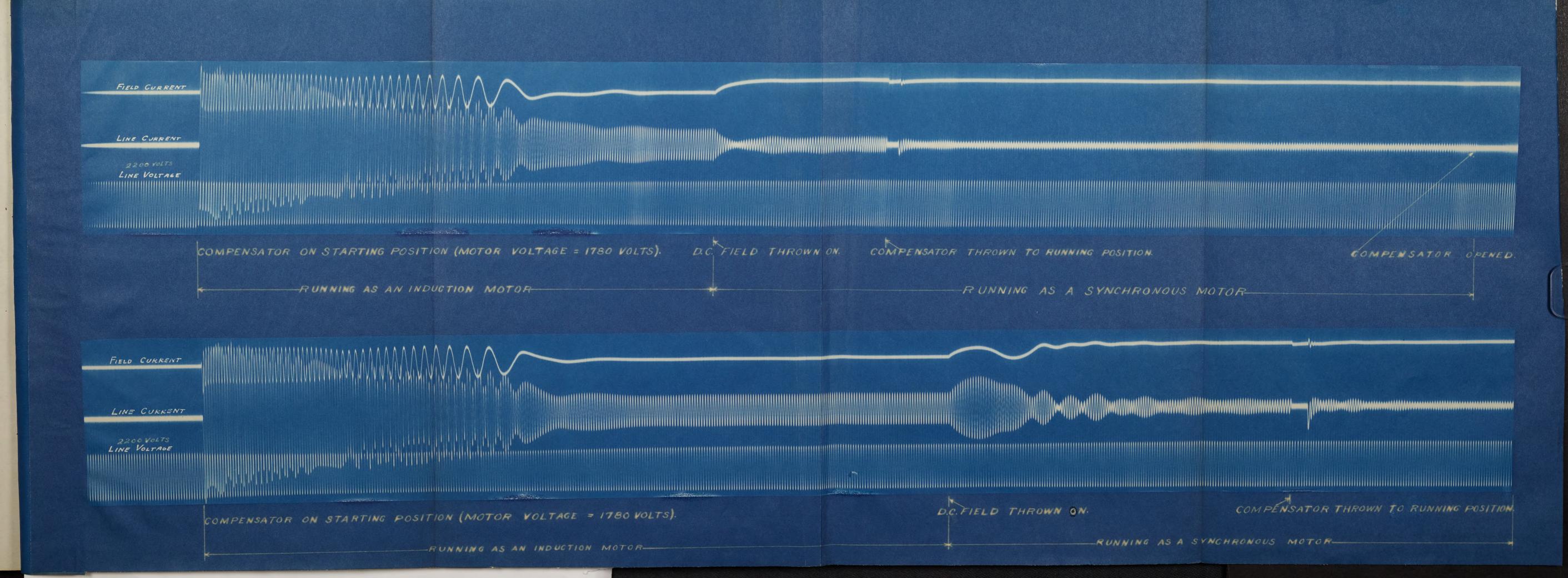
- A. Saturation Curve.
- B. Synchronous Impedance Curve.



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